

Xiao Liu 1,2, Qinghua Zeng 2,[*](https://orcid.org/0000-0002-5529-1031) , Zhengzhi Ding ² and Haitao Xu [3](https://orcid.org/0000-0002-3664-3097)

- ¹ School of Electronic, Electrical and Communication Engineering, University of Chinese Academy of Sciences, Beijing 100049, China
- ² Aerospace Information Research Institute, Chinese Academy of Sciences, Beijing 100094, China
³ Department of Communication Engineering, University of Science and Technology Beijing
- ³ Department of Communication Engineering, University of Science and Technology Beijing,
- Beijing 100083, China ***** Correspondence: zengqh@aircas.ac.cn

Abstract: Ku-band drive amplification and a 7-bit true-time-delay (TTD) function were realized as a part of a LTCC-based T/R module to increase integration. The 8-channel T/R module was fabricated and its key characteristics were measured, including a 3-bit $(1/2/4 \lambda)$ TTD, 4-bit $(0.25/0.5/1/2 \lambda)$ TTD, receive gain, noise figure and output power. The 8-channel T/R module can be further adopted to increase bandwidth and scanning angle of phased arrays without beam squint.

Keywords: Ku-band; T/R module; true-time-delay; phased array

1. Introduction

Active electronically scanned arrays (AESA) with a number of transmit/receive (T/R) modules are widely adopted due to the rapid change of beam shape and beam direction [\[1](#page-8-0)[–3\]](#page-8-1). Phase shifters are used to adjust the phase of each element and yet beam squint still occurs for wide-band applications [\[4–](#page-8-2)[7\]](#page-9-0). Therefore, true-time-delay (TTD) modules are essential in phased array systems such as AESA radars to meet the increasing demand of broader bandwidth and wider scanning angle [\[8](#page-9-1)[–11\]](#page-9-2).

Fenn et al., reported a 300–450 MHz dipole linear array antenna and TTDs can be implemented by using appropriate lengths of coaxial cable transmission lines in order to maintain the main beam pointed at a fixed 40-degree angle from broadside over the operating bandwidth [\[12\]](#page-9-3). However, more and more functions are designed to be combined with TTD function to meet the increasing demand of higher integration density and smaller module size. Liu et al., reported an X-band receiver integrated with a subarray-level 5-bit TTD module [\[13\]](#page-9-4). Shi et al., reported a Ka-band high integration receiver module which combines TTD units at sub-array level and phase shifters at array-element level to realize a 1.6 GHz instantaneous bandwidth [\[14\]](#page-9-5). Tang et al. reported an X-band 5-bit timedelay amplifier module in which a time-delay function is realized by coplanar waveguide (CPW) transmission lines [\[15\]](#page-9-6). Guo et al. reported a 6-bit TTD amplifier module for X-band subarray phased arrays where CPW transmission lines are also used as time delay units [\[16\]](#page-9-7). Li et al., reported a C-band subarray drive time-delay module which integrates 5-bit timedelay, amplification and power dividing/combining [\[17](#page-9-8)[,18\]](#page-9-9). Li et al., reported an X-band subarray drive time-delay module which integrates 5-bit time-delay, and receives and transmits gain compensation [\[19\]](#page-9-10). In both modules, microstrip lines and striplines with different lengths are used to realize time-delay function. However, using CPW transmission lines or microstrip lines and striplines as delay paths results in an unacceptable amount of module area for the required time delay, and the insertion loss of different time delay states varies substantially [\[20](#page-9-11)[,21\]](#page-9-12).

In terms of electrical design and micro/macro-packaging fabrication, electronic delay lines using monolithic microwave integrated circuit (MMIC) devices can be more easily

Citation: Liu, X.; Zeng, Q.; Ding, Z.; Xu, H. A LTCC-Based Ku-Band 8-Channel T/R Module Integrated with Drive Amplification and 7-Bit True-Time-Delay. *Sensors* **2022**, *22*, 6568. [https://doi.org/10.3390/](https://doi.org/10.3390/s22176568) [s22176568](https://doi.org/10.3390/s22176568)

Academic Editor: Juan Fang

Received: 8 July 2022 Accepted: 25 August 2022 Published: 31 August 2022

Publisher's Note: MDPI stays neutral with regard to jurisdictional claims in published maps and institutional affiliations.

Copyright: © 2022 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license [\(https://](https://creativecommons.org/licenses/by/4.0/) [creativecommons.org/licenses/by/](https://creativecommons.org/licenses/by/4.0/) $4.0/$).

implemented through multi-chip module (MCM) technology for higher integration and smaller circuit diagrams [\[22–](#page-9-13)[25\]](#page-9-14). Zhang et al., reported a C-band multi-channel 9-bit TTD amplification module, and GaAs MMICs are adopted to realize TTD function by switching $\frac{m}{p}$ module). The data can be hand control at a model to realize TTD function by switching between a reference path and a delay path [\[26](#page-9-15)[,27\]](#page-9-16). However, on the one hand, extra bi-Edirectional amplifiers are needed to compensate the signal transmission loss induced by insertion loss of TTD MMICs [\[28\]](#page-9-17). Not only TTD units but also amplifiers are integrated in
by insertion loss of TTD MMICs [28]. Not only TTD units but also amplifiers are integrated in one active MMIC device to combat insertion losses and to provide gain [\[29,](#page-9-18)[30\]](#page-9-19). On the other hand, it is barely reported that TTD modules and T/R modules are integrated as a whole.
Little integrated as a whole Integrating a TTD module with a T/R module can greatly improve the packaging density and simplify the interconnection among modules inside a complicated antenna system. density and simplify the interconnection among modules inside a complicated antenna μ -directional amplifiers are needed to compensate the signal transmission loss induced to compute the signal transmission loss induced

In this paper, a LTCC-based Ku-band TTD module is integrated with a Ku-band T/R module inside the same shell. LTCC technology is very suitable for the manufacturing of mixed-signal multi-layer substrates, which can realize the combined design of microwave c ircuits, logi $\rm c$ control circuits and power management circuits on different layers. The TTD module integrates a 7-bit time-delay, receive gain compensation and transmit power drive, which are realized through active MMICs.

The paper is organized as follows: the topography and link budget of the 8-channel T/R module with 7-bit TTD are presented in Section 2. Section 3 reports the simulation results of the SMP-microstrip vertical transition and strip line power divider for the Kuband. Measurements of the 8-channel T/R module with 7-bit TTD are reported in Section $4\,$ $4\,$ and conclusions are provided in Section 5.

2. Topography and Link Budget of the 8-Channel T/R Module with 7-Bit TTD

As shown in Figures 1 and [2,](#page-2-0) a 4-bit $(0.25/0.5/1/2 \lambda)$ TTD MMIC $(f_0 = 16.7 \text{ GHz})$ and *t*_{delay} = 15 ps) is used for every four channels and a 3-bit (1/2/4 λ) TTD MMIC $(f_0 = 16.7 \text{ GHz}$ and $t_{\text{delay}} = 60 \text{ ps}$) is used for every eight channels. Not only TTD function but also receive gain compensation*,* transmit power drive and switches are all integrated in TTD MMICs.

Figure 1. Topography of the 8-channel T/R module with 7-bit TTD.

Each channel mainly consists of a GaAs beamformer MMIC, a GaAs low noise amplifier (LNA), a GaAs limiter, a GaN power amplifier (PA), a circulator, a pulse modulation circuit and a power supply circuit. Amplifiers, phase shifters, attenuators and serial-toparallel converters are integrated in the beamformer MMIC. A 2-way power divider and two 4-way power dividers are realized through LTCC-based striplines. The circulators used in this module are three-port ferrite circulators. Under transmit mode, a wave incident on port 1 is coupled into port 2; under receive mode, a wave incident on port 2 is coupled into port 3. The typical isolation S32 is more than 27 dB.

Figure 2. Link budget of receive (red) and transmit (blue). **Figure 2.** Link budget of receive (red) and transmit (blue).

Under receive mode, the theoretical net gain is 41 dB and the noise figure (NF) is less than 3.6 dB. Under transmit mode, the maximum output power is 42.1 dBm (16 W). incident on port 1 is coupled into port 2; under receive mode, a wave incident on port 2 is Under receive mode, the theoretical net gain is 41 dB

and two 4-way power dividers are realized through LTCC-based striplines. The circula-based striplines

Increased integration density and decreased weight are achieved with 15 layers of stacked LTCC substrates and a 50AlSi shell. High power GaN PAs are welded to molybdenum copper substrates and then sintered on 50AlSi shells to improve heat dissipation. The fab-ricated 8-channel T/R module is shown in Figure [3.](#page-3-1) The size is 110 mm \times 65 mm \times 10 mm and the weight is 85 g. 10 mm and the weight is 85 g. $T_{\rm eff}$ module is shown in Figure 3. The size is shown in Figure 3. The size is 110 mm \sim 65 mm \sim 65 mm \sim

Figure 3. A photo of the 8-channel T/R module with 7-bit TTD. **Figure 3.** A photo of the 8-channel T/R module with 7-bit TTD.

3. Simulation of SMP-Microstrip Vertical Transition and Stripline Power Divider for 3. Simulation of SMP-Microstrip Vertical Transition and Stripline Power Divider Ku-Band for Ku-Band

In this module, a vertical transition of SMP-to-microstrip is adopted, as illustrated in Figure [4.](#page-3-2) The SMP pin passes through the CLTE-XT substrate with a metallized via. Simulation was carried out to optimize the three critical parameters: the radius of the lized via, the radius of the annular ring and the width between the annular ring and the width between the annul metallized via, the radius of the annular ring and the width between the annular ring and the ground, and the width of the microstrip line.

Figure 4. SMP-to-microstrip vertical transition model: (a) front side of the simulation model; (b) back $\frac{1}{1}$ side of the simulation model.

Simulation results of S11, S22 and S21 are shown in Figure [5.](#page-4-0) The return loss is below −25 dB and the insertion loss is below −0.15 dB. −25 dB and the insertion loss is below −0.15 dB.

Figure 5. Simulation results of the SMP-to-microstrip vertical transition. (a) Return loss S11 and S22; (**b**) Insertion loss S21. (**b**) Insertion loss S21.

LTCC substrates are adopted for the T/R modules and TTD modules. A 2-way Wilkinson power divider and a 4-way Wilkinson power divider are realized by means of \overline{MSG} LTCC-based stripline to further increase integration. The simulation model of the LTCC-LTCC-based stripline 2-way power divider is shown in Figure [6.](#page-4-1) Alongside the stripline, grounding vias are adopted to constrain the transmission of the electromagnetic wave to decrease the loss of electromagnetic energy. Based on the 2-way power divider, the 4-way power divider is realized through a cascading two-level 2-way power divider. of LTCC-based stripline to further increase integration. The simulation model of the

Figure 6. LTCC-based stripline 2-way power divider for Ku-band. **Figure 6.** LTCC-based stripline 2-way power divider for Ku-band.

The simulation results are shown in Figure [7.](#page-5-1) The insertion loss is below -3.5 dB the difference between that of port-3 is less than $\frac{1}{2}$ is less than $\frac{1}{2}$ and the difference between that of port-2 and port-3 is less than 0.1 dB. The return loss of port-1 is less than −20 dB and the return loss of port-2 and port-3 is less than −30 dB. The isolation between port-2 and port-3 is less than −20 dB.

(**c**)

Figure 7. Simulation results of the LTCC-based stripline 2-way power divider: (a) insertion loss S21 and S31; (**b**) return loss S11, S22 and S33; (**c**) isolation S23. and S31; (**b**) return loss S11, S22 and S33; (**c**) isolation S23.

4. Measurement of the 8-Channel T/R Module with 7-Bit TTD 4. Measurement of the 8-Channel T/R Module with 7-Bit TTD

Measurement of the 8-channel T/R module with 7-bit TTD includes TTD, NF, receive Measurement of the 8-channel T/R module with 7-bit TTD includes TTD, NF, receive gain and output power. gain and output power.

4.1. TTD 4.1. TTD

The TTD units in active MMICs are basically composed of SPDT switches, DPDT switches and time delay elements such as CLC $\overline{\pi}$ -networks. TTD is calculated by the wrapped phase, which was measured through Agilent vector network analyzer N5244A. unwrapped phase, which was measured through Agilent vector network analyzer N5244A.

Firstly, zero-centering is carried out based on the unwrapped phase of the middle Firstly, zero-centering is carried out based on the unwrapped phase of the middle frequency, as shown in Equation (1). frequency, as shown in Equation (1).

$$
P(f) = P(f_x) - P(f_0),
$$
\n(1)

 $P(f_0)$ is the unwrapped phase of the middle frequency and $P(f_x)$ is the unwrapped phase of each frequency within the bandwidth.

Secondly, the delay wavelength of each frequency n_x is calculated based on the middle $(2).$ frequency, as shown in Equation (2).

$$
n_x = P(f) / [360 \times (1 - f_x / f_0)], \tag{2}
$$

Thirdly, the least square method is adopted to fit n_x and the fitted n is further used to calculate the TTD value $\Delta P(f)$, as shown in Equation (3).

$$
\Delta P(f) = P(f) - n \times 360 \times (1 - f_x/f_0),\tag{3}
$$

Measured TTD and the error with a theoretical value are listed in Table [1.](#page-6-0) For the 4-bit TTD, the maximum relative error is approximately 15%. For the 3-bit TTD, the maximum **relative error is approximately 10%.**

Theoretical TTD/λ	Measured TTD/λ	Absolute Error/ λ	Relative Error/%
0.25 (of 4-bit TTD)	0.2880	0.0380	15.20%
0.5 (of 4-bit TTD)	0.4957	0.0043	0.86%
1 (of 4-bit TTD)	0.9565	0.0435	4.35%
2 (of 4-bit TTD)	2.0057	0.0057	0.29%
1 (of 3-bit TTD)	0.8954	0.1046	10.46%
2 (of 3-bit TTD)	2.0998	0.0998	4.99%
4 (of 3-bit TTD)	4.1074	0.1074	2.69%

Table 1. Measurement results of 7-bit TTD and accuracy of each state. **TADIC 1.** MCasulCHICHI TCSUNS OF 7 DIT I TD and accuracy of cach state.

4.2. Noise Figure 8, the measured results of NF are below 3.65 dB with the bandwidth of 1.2 GHz, the bandwidth of 1.

NF was measured using the Agilent Noise Figure Analyzer N8975A. As shown in Figure [8,](#page-6-1) the measured results of NF are below 3.65 dB within the bandwidth of 1.2 GHz, which meets the theoretical value of 3.6 dB.

Figure 8. Measured results of NF.

Figure 8. Measured results of NF. *4.3. Receive Gain*

The characteristics of receive were measured using the Agilent N5244A Network Analyzer. SWR within the bandwidth (from 16.1 GHz to 17.3 GHz) is less than 1.4. The standing wave ratio (SWR) of the receive input port is shown in Figure [9.](#page-7-0) The

The minimum output power of the network analyzer is -30 dBm and a 10 dB attenuator is added in front of the receive input port so that the whole module works in the linear region. The measurements of receive within the bandwid[th o](#page-7-1)f 1.2 GHz are shown in Figure 10. S21 is approximately 22 dB with a difference within the bandwidth of less than 0.5 dB. The net gain of receive is 22 + 10 (10 dB attenuator) + 9 (4-way and 2-way divider) = 41 dB.

4.4. Output Power 4.4. Output Power 4.4. Output Power

Under transmit mode, the output power was measured using the Agilent N1912A Under transmit mode, the output power was measured using the Agilent N1912A Power Meter. Within the bandwidth of 1.2 GHz, GaN power amplifiers can work under saturation mode with input power varying from -2 dBm to 2 dBm. The maximum duration for transmit is 100 microseconds with a duty ratio of 10%. The measured output power is α larger than 42.3 dBm (17 W). power is larger than 42.3 dBm (17 W).

The measurements of transmit through the network analyzer are shown in Figure [11.](#page-8-4) A 30 dB attenuator is added in front of port 2 to protect the network analyzer from overpower. The insertion loss of the cable for measurement is about 3 dB.

Figure 11. Measured results of transmit. **Figure 11.** Measured results of transmit.

5. Conclusions The measured output power is between 9.12 dBm and 9.76 dBm, with a difference of A and B , B and B module with σ function was designed and fabricless than 0.7 dB.

cated. LTCC-integrated power dividers and vertical transmission of SMP-to-microstrip **5. Conclusions**

 \mathcal{M} were designed and simulated. Each state of the 7-bit TTD was measured and the relative relative relative \mathcal{M} A Ku-band 8-channel T/R module with 7-bit TTD function was designed and fabri-
A LIECC interesting the difference of the difference eated. EFCC-integrated power dividers and vertical transmission of 5MF-to-interestip-
were designed and simulated. Each state of the 7-bit TTD was measured and the relative error was less than 15.2%. Under receive mode, the net gain was 41 dB with a difference error was less than 15.2%. Under receive mode, the net gain was 41 dB with a difference within the bandwidth of 1.2 GHz of less than 0.5 dB. The measured NF was less than 3.65 dB. Under transmit mode, the output power was larger than 42.3 dBm (17 W) with $\frac{1}{2}$. a difference of less than 0.7 dB. cated. LTCC-integrated power dividers and vertical transmission of SMP-to-microstrip

have read and agreed to the published version of the manuscript. writing—original draft preparation, X.L.; writing—review and editing, Q.Z. and H.X. All authors have read and agreed to the published version of the manuscript. **Author Contributions:** Methodology, X.L.; software, Q.Z.; formal analysis, Z.D.; investigation, X.L.;

Funding: This research received no external funding.

Data Availability Statement: Not applicable. **Institutional Review Board Statement:** Not applicable.

Informed Consent Statement: Not applicable.

Data Availability Statement: Not applicable.

1. Mailloux, R.J. *Phased Array Antenna Handbook*, 2nd ed.; Artech House: Boston, MA, USA, 2005. **Conflicts of Interest:** The authors declare no conflict of interest.

$2.$ References **References**

- Proceedings of the IEEE Wireless Communications and Networking Conference Workshops, Doha, Qatar, 3–6 April 2016; p. 1. Mailloux, R.J. *Phased Array Antenna Handbook,* 2nd ed.; Artech House: Boston, MA, USA, 2005.
2. Participal article in Termina Termina Termina and Termina and Termina and Termina and Termina and Termina a
- 2. Orakwue, S.I.; Ngah, R.; Rahman, T.A. A two dimensional beam scanning array antenna for 5G wireless communications. In Proceedings of the IEEE Wireless Communications and Networking Conference Workshops, Doha, Qatar, 3–6 April 2016; p. 433.
Isonalist 2016 $[CDSSNet]$
 $[20.10931C1]$ [\[CrossRef\]](http://doi.org/10.1109/WCNCW.2016.7552738)
- 3. Das, A.; Puri, M.; Sengar, J.S. A novel monolithic integrated phased array antenna using 4-bit distributed MEMS phase shifter and triangular patch antenna. In Proceedings of the International Conference on ICACCI, Mysore, India, 22–25 August 2013; p. 913.
Is $\frac{1}{20}$ $\frac{1}{20}$. $\frac{1}{20}$ $\frac{1}{20}$ $\frac{1}{20}$ $\frac{1}{20}$ $\frac{1}{20}$ $\frac{1}{20}$ [\[CrossRef\]](http://doi.org/10.1109/ICACCI.2013.6637298)
- 4. Chen, B.; Yang, M.; Wang, Y.; Dang, X.; Wu, B. The applications and future of synthetic impulse and aperture radar. In Proceedings
6.1. 2016.OFF J. (2016) of the 2016 CIE International Conference on Radar (RADAR), Guangzhou, China, 10–13 October 2016; p. 1. [\[CrossRef\]](http://doi.org/10.1109/RADAR.2016.8059190)
Dealer Machine 1986, the 2018 to the context of the context of the context of the context of the 2018 to the
- 5. Duan, M.; Ma, J. Wide range and high resolution true time delay unit for phased array antenna. In Proceedings of the 2018 Asia
Case of the 2018 Asia Communications and Photonics Conference (ACP), Hangzhou, China, 26–29 October 2018. [\[CrossRef\]](http://doi.org/10.1109/ACP.2018.8596111)
- 6. Guntupalli, A.B.; Djerafi, T.; Wu, K. Two-dimensional scanning antenna array driven by integrated waveguide phase shifter. *IEEE Trans. Antennas Propag.* **2014**, *62*, 1117. [\[CrossRef\]](http://doi.org/10.1109/TAP.2013.2292935)
- 7. Yeo, W.-G.; Nahar, N.K.; Sertel, K. Phased array antenna with integrated MEMS phase shifters for Ka-band SATCOM. In Proceedings of the IEEE Antennas and Propagation Society International Symposium, Orlando, FL, USA, 7–13 July 2013; Volume 105. [\[CrossRef\]](http://doi.org/10.1109/APS.2013.6710713)
- 8. Rotman, R.; Tur, M.; Yaron, L. True Time Delay in Phased Arrays. *Proc. IEEE* **2016**, *104*, 504–518. [\[CrossRef\]](http://doi.org/10.1109/JPROC.2016.2515122)
- 9. Tong, D.T.K.; Wu, M.C. Transmit/receive module of multiwavelength optically controlled phased-array antennas. *IEEE Photonics Technol. Lett.* **1998**, *10*, 1018–1020. [\[CrossRef\]](http://doi.org/10.1109/68.681303)
- 10. Takahashi, T.; Konishi, Y.; Hariu, K.; Nakaguro, H.; Chiba, I. Beam pointing error analysis for phased array antennas with true time delay modules. In Proceedings of the IEEE Antennas and Propagation Society International Symposium, San Antonio, TX, USA, 16–21 June 2002; Volume 1, pp. 606–609. [\[CrossRef\]](http://doi.org/10.1109/APS.2002.1016418)
- 11. Chen, Y.; Chen, R.T. A fully packaged true time delay module for a K-band phased array antenna system demonstration. *Proc. IEEE Photonics Technol. Lett.* **2002**, *14*, 1175–1177. [\[CrossRef\]](http://doi.org/10.1109/LPT.2002.1022009)
- 12. Fenn, A.J.; Hurst, P.T.; Parad, L.I.; Bruno, D.M. Ultrawideband time-delay steered UHF dipole linear array antenna. In Proceedings of the IEEE International Symposium on Phased Array Systems and Technology, Boston, MA, USA, 15–18 October 2013; pp. 731–737.
- 13. Liu, W.; Wang, Y. Study of the true-time-delay microwave module technology. *Sci. Technol. Innov.* **2015**, *18*, 79–80.
- 14. Shi, H.; Zhang, L.; Fan, Z.; Xue, X.; Zhang, T.; Zhou, D. Design of a Ka Band High Integration Receiver Module. In Proceedings of the International Conference on Communication Technology, Bandung, Indonesia, 3–4 May 2018.
- 15. Tang, Y.Y.; Wang, W.; Li, C.M.; Song, X.D. Design of a CPW time-delay amplifier module. *Telecommun. Inf.* **2016**, *538*, 38–40.
- 16. Guo, Y.; Shang, C.; Liu, K.; Wang, L.; Liu, X.; Xu, Y.; Zhang, T. A True-Time-Delay Transmit/ Receive Module for X-band Subarray Phased Arrays. *IEICE Electron. Express* **2017**, *14*, 20171039. [\[CrossRef\]](http://doi.org/10.1587/elex.14.20171039)
- 17. Li, S.; Wang, X.; Wang, Q. Miniaturization and high accuracy design of a C-band drive time-delay module. *J. Microw.* **2016**, *32*, 78–81+87.
- 18. Shu-liang, L.; Qi, W.; Hui, W. Research on electromagnetic compatibility of miniaturized drive time-delay module. In Proceedings of the IEEE International Radar Conference, Philadelphia, PA, USA, 2–6 May 2016; pp. 1–4.
- 19. Li, S.; Zhu, R.; Liu, Y. Design and realization of an X-band subarray drive time-delay module for active phased array radar. *Mod. Radar* **2016**, *38*, 52–54+71.
- 20. Ge, H.; Liu, Y.; He, J. Amplitude compensation design of a C-band time delay module. *Radar ECM* **2017**, *37*, 15–18.
- 21. Ulusoy, A.; Schleicher, B.; Schumacher, H. A tunable differential all-pass filter for UWB true time delay and phase shift applications. *IEEE Microw. Compon. Lett.* **2011**, *21*, 462. [\[CrossRef\]](http://doi.org/10.1109/LMWC.2011.2162613)
- 22. Hu, F.; Mouthaan, K. A 1–21 GHz, 3-bit CMOS true time delay chain with 274 ps delay for ultra-broadband phased array antennas. In Proceedings of the IEEE Radar Conference, Johannesburg, South Africa, 27–30 October 2015. [\[CrossRef\]](http://doi.org/10.1109/EuRAD.2015.7346303)
- 23. Hu, F.; Mouthaan, K. A 1–20 GHz 400 ps true-time delay with small delay error in 0.13 µm CMOS for broadband phased array antennas. In Proceedings of the IEEE Microwave Symposium, Phoenix, AZ, USA, 17–22 May 2015. [\[CrossRef\]](http://doi.org/10.1109/MWSYM.2015.7166834)
- 24. Lin, W.; Li, S. Realization of a miniaturized millimeter-wave time-delay module. *Mod. Radar.* **2020**, *42*, 64–66+70.
- 25. Kim, M. A stepped-impedance true time delay line using GaAs MMIC technology. *IEICE Electron. Express* **2016**, *13*, 20160463. [\[CrossRef\]](http://doi.org/10.1587/elex.13.20160463)
- 26. Zhang, L.; Li, Z.; Chen, K.; Chen, G.; Zhao, Y. Compact Size Multi-Channel True Time Delay Module with High Accuracy. In Proceedings of the 2018 China International SAR Symposium, CISS, Shanghai, China, 10–12 October 2018.
- 27. Bettidi, A.; Carosi, D.; Corsaro, F.; Marescialli, L.; Nanni, A.; Romanini, P. MMIC Chipset for wideband multifunction T/R Module. IEEE MTT-S International Microwave Symposium Digest. In Proceedings of the IEEE MTT-S International Microwave Symposium, Baltimore, MD, USA, 5–10 June 2011; pp. 1–4. [\[CrossRef\]](http://doi.org/10.1109/MWSYM.2011.5973298)
- 28. Longbrake, M. True time-delay beamsteering for radar. In Proceedings of the 2012 IEEE National Aerospace and Electronics Conference (NAECON), Dayton, OH, USA, 25–27 July 2012; pp. 246–249. [\[CrossRef\]](http://doi.org/10.1109/NAECON.2012.6531062)
- 29. Cho, M.-K.; Han, J.-H.; Kim, J.-H.; Kim, J.-G. An X/Ku-band bi-directional true time delay T/R chipset in 0.13 µm CMOS technology. In Proceedings of the 2014 IEEE MTT-S International Microwave Symposium (IMS2014), Tampa, FL, USA, 1–6 June 2014; pp. 1–3. [\[CrossRef\]](http://doi.org/10.1109/MWSYM.2014.6848608)
- 30. Cho, M.-K.; Yoon, S.-H.; Sim, S.; Jeon, L.; Kim, J.-G. CMOS-based Bi-directional T/R chipsets for phased array antenna. In Proceedings of the 2012 IEEE/MTT-S International Microwave Symposium Digest, Montreal, QC, Canada, 17–22 June 2012; pp. 1–3. [\[CrossRef\]](http://doi.org/10.1109/MWSYM.2012.6259562)