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# Simulation Study: The Impact of Structural Variations on the Characteristics of a Buried-Channel-Array Transistor (BCAT) in DRAM

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Abstract: As the physical dimensions of cell transistors in dynamic random-access memory (DRAM) have been aggressively scaled down, buried-channel-array transistors (BCATs) have been adopted in industry to suppress short channel effects and to achieve a better performance. In very aggressively scaled-down BCATs, the impact of structural variations on the electrical characteristics can be more significant than expected. Using a technology computer-aided design (TCAD) tool, the structural variations in BCAT (e.g., the aspect ratio of the BCAT recess-to-gate length, BCAT depth, junction depth, fin width, and fin fillet radius) were simulated to enable a quantitative understanding of its impact on the device characteristics, such as the input/output characteristics, threshold voltage, subthreshold swing, on-/off-current ratio, and drain-induced barrier lowering. This work paves the road for the design of a variation-robust BCAT.

**Keywords:** buried channel array transistor; structural variation; short channel effects; device characteristics

## 1. Introduction

The physical dimensions of transistors in integrated circuit (IC) have been scaled down (i) to make the density of the transistors in ICs as high as possible and (ii) to improve the electrical performance of the transistors [1,2]. However, as the channel length of the transistors (including the metal oxide semiconductor field effect transistor (MOSFET)) has been aggressively decreased, and the short channel effect (SCE) has adversely affected the devices' performance. Note that the threshold voltage in MOSFET becomes lower with a shorter channel length (which is an undesirable secondary effect). To address the SCE issue, many engineering solutions have been proposed, such as (i) the use of a new device architecture to enhance the gate-to-channel capacitive coupling (e.g., fin-shaped FET, ultra-thin-body FET, and multiple-bridge-channel FET), (ii) the use of novel materials (e.g., SiGe in source/drain or compound semiconductors in the channel to induce appropriate stress, resulting in better mobility of the electrons/holes), and (iii) the high-k/metal-gate technique for achieving an electrically thin but physically thick gate oxide layer [3–8]. One of the solutions for overcoming the SCE in DRAM (dynamic random-access memory) cell transistors is to adopt the buried-channel-array transistor (BCAT) structure. This device structure can increase the effective channel length due primarily to its recessed channel and buried gate structure [9]. However, as the 3D physical dimensions of BCATs have been significantly scaled down, it is important to study the impact of structural variations on the electrical characteristics of BCATs. Moreover, process-induced systematic/random variation results in undesired alterations to the device characteristics [10]. Thus, the impact of structural variations on the electrical characteristics of BCATs must be quantitatively



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**Copyright:** © 2022 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https:// creativecommons.org/licenses/by/ 4.0/). studied in order to control device characteristics in mass production and pave the road for the design of next-generation BCAT devices. In this work, a BCAT device with 20 nm-long gate length was simulated with a 3D technology computer-aided design (TCAD) tool (i.e., Synopsys Sentaurus). To obtain a nominal device structure, its device design parameters were altered in order to reveal the ways in which each parameter's variation affected the device characteristics.

#### 2. Baseline Device Design and Its Structural Variations

The BCAT device was built with the Sentaurus TCAD tool, and a 3D bird's eye view of the device is shown in Figure 1a. The saddle-fin-shaped silicon channel was buried under the nitride (Si<sub>3</sub>Ni<sub>4</sub>) insulator layer and covered by the tungsten gate and silicon oxide (see Figure 1b). The cross-sectional views across the channel and along the channel is shown in Figure 1c,d, respectively. The physical gate length (Lgate) and recess (Drecess) of the baseline BCAT device was nominally set to 20 nm and 120 nm, respectively, resulting in a  $D_{\text{recess}}/L_{\text{gate}}$  (= AR<sub>gate</sub>) of ~ 6.0 (see Figure 2a). Its gate material was tungsten, with the working function of 4.8 eV. The recessed region and the saddle fin of the baseline BCAT device were surrounded by the gate oxide of 5 nm. Note that the D<sub>BCAT</sub> corresponded to the thickness of the nitride in gate stack (see Figure 2b). The silicon substrate/body region was doped with 10<sup>17</sup> cm<sup>-3</sup> boron, while the source and drain regions were counter-doped with  $10^{20}$  cm<sup>-3</sup> arsenic. Note that the Gaussian doping profile was used for the device. The nominal junction depth ( $D_{junction}$ ) was set to 40% of  $D_{recess}$  (see Figure 2c vs. Figure 2a). The saddle fin width ( $W_{fin}$ ) of the nominal BCAT device was set to 17 nm (see Figure 2d). The saddle fin fillet radius (R<sub>fillet</sub>) was defined as the multiplication factor of the saddle fin radius (see Figure 2e). This quantitatively indicated whether the saddle fin shape was rounded or angled [11]. The nominal R<sub>fillet</sub> was set to 1.0 to ensure that the saddle fin was shaped as a semi-circle.



**Figure 1.** (a) A 3D bird's eye view of the buried-channel-array transistor (BCAT), (b) its saddle-finshaped silicon channel, and its cross-sectional view (c) across the channel (A–A') and (d) along the channel (B–B'). Note that the substrate/body region was doped with  $10^{17}$  cm<sup>-3</sup> boron, while the source and drain regions were counter-doped with  $10^{20}$  cm<sup>-3</sup> arsenic.



**Figure 2.** Parametric variations in the BCAT: (**a**) aspect ratio of the BCAT recess ( $D_{recess}$ )-to-gate length ( $L_{gate}$ ) (i.e.,  $AR_{Gate} = D_{recess}/L_{gate}$ ), (**b**) BCAT depth ( $D_{BCAT}$ ), (**c**) junction depth ( $D_{junction}$ ), (**d**) fin width ( $W_{fin}$ ), and (**e**) fin fillet radius ( $R_{fillet}$ ).

The impacts of structural variations in baseline device on its electrical characteristics were investigated. The aspect ratio of  $D_{recess}$  to  $L_{gate}$  (i.e.,  $AR_{gate}$ ) varied from 5 (-17% of the baseline) up to 7 (+17% of the baseline). Note that the  $AR_{gate}$  of the baseline device structure was 6. The BCAT depth ( $D_{BCAT}$ ) varied from 24 nm (-33% of the baseline) to 48 nm (+33% of the baseline). Note that the  $D_{BCAT}$  of the baseline device was 36 nm. The  $D_{junction}$  (which is defined as the depth at which the doping concentration is  $10^{17}$  cm<sup>-3</sup>) varied from 30% to 50% of the  $D_{recess}$ . The saddle fin width ( $W_{fin}$ ) varied from 11 nm to 23 nm (i.e.,  $-33 \sim +33\%$  of the baseline value). Note that the nominal saddle fin width ( $W_{fin}$ ) of the baseline device was 17 nm. The saddle fin fillet radius ( $R_{fillet}$ ) of the baseline). For each structural variation mentioned above, the input/output characteristics (i.e.,  $I_D$ -vs. -V<sub>G</sub>/ $I_D$ -vs.-V<sub>D</sub>) were simulated (see Table 1).



**Table 1.**  $I_D$ - $V_G$  and  $I_D$ - $V_D$  of the BCAT for given parametric variations, including AR<sub>gate</sub>,  $D_{BCAT}$ ,  $D_{junction}$ ,  $W_{fin}$ , and  $R_{fillet}$ . Note that the drain current is normalized to the channel width.

The Philips unified mobility model was adopted for the simulations to assess the dependence of the mobility on the electron-hole scatterings, screening of ionized impurities by charged carriers, and clustering of impurities [12]. The mobility degradation at the semiconductor–insulator interfaces due to surface roughness scattering was incorporated into in the simulations using the Lombardi mobility model [13]. The Canali model was taken into account for the carrier velocity saturation in the regions with high and low electric fields [14]. The Hurkx trap-assisted tunnelling model was used to observe the band-to-band tunnelling [15].

### 3. Results and Discussion

For the given parameters, including  $AR_{gate}$ ,  $D_{BCAT}$ ,  $D_{junction}$ ,  $W_{fin}$ , and  $R_{fillet}$ , the input/output characteristics (i.e.,  $I_D$ -vs.- $V_G/I_D$ -vs.- $V_D$ ) of the baseline device were simulated with the structural variations. The results are summarized in Table 1.

It is noteworthy that the short channel effects of the DRAM cell transistor in the  $\sim 20$  nm technology mode were effectively rebuilt (see the  $I_D$  vs.  $V_D$  plots in Table 1). From the  $I_D$  vs. V<sub>G</sub> plots, key device characteristics were extracted, and they are summarized in Figure 3a–d. Herein, the threshold voltage (Vth) was defined using the constant current method (i.e., the constant current =  $10^{-7}$  A × W (channel width)/L (channel length) [16]). The nominal (baseline structure) device performance metrics were:  $V_{th} = 0.656 \text{ V}$ ; SS = 76 mV/dec; on-/off-current ratio =  $3.4 \times 10^{10}$ ; and DIBL = 23.6 mV/V. Note that these results are within reasonable ranges  $(V_{th} \approx 0.7 \text{ V}; \text{SS} < 90 \text{ mV/dec}; \text{ on-/off-current ratio} \approx 10^{10}; \text{DIBL} \approx <50 \text{ mV/V})$  compared to the other reported/published experimental results, such as those of the 20 nm FinFET device or 30 nm buried-word-line-structured device [17, 18]. Higher AR<sub>gate</sub> induced a lower V<sub>th</sub>, steeper subthreshold swing (SS), better drain-induced barrier lowering (DIBL), and a higher on-/offcurrent ratio. This is due primarily to the higher height of the saddle fin, as well as the channel region being more closely surrounded by the gate, resulting in a higher gate-to-channel coupling capacitance. Note that increasing  $AR_{gate}$  significantly affected the SS (i.e., from 74.1 mV/dec to 78.6 mV/dec; nominal SS = 76.0 mV/dec) and on-/off-current ratio (i.e., from  $2.2 \times 10^9$ to  $6.2 \times 10^{11}$ ; nominal on-/off-current ratio =  $3.4 \times 10^{10}$ ) at  $\pm 17\%$  of the baseline. Though increasing the AR<sub>gate</sub> yielded a better performance in terms of the device characteristics (i.e., a steeper SS, higher on-/off-current ratio, and lower DIBL), the difficulties involved in the fabrication process must be considered. To achieve the higher AR<sub>gate</sub>, a deeper BCAT recess (D<sub>recess</sub>) at the same gate length is needed, requiring more advanced technologies in the lithography and etching processes. In addition, the increase in AR<sub>gate</sub> must be limited to the marginal point in order to avoid the risks of the bending/leaning of the Si active substrate or voids in the gate materials (tungsten, in this study) during the deposition process (due to imperfect deposition in the deep BCAT recessed area) [19,20]. Increasing the  $D_{BCAT}$  resulted in a lower  $V_{th}$ , non-steeper SS, lower on-/off-current ratio, and a worse DIBL. This is mainly because (1) the gate controllability decreased as the effective gate length became shorter (herein, the effective gate length was defined as the distance from the bottom of the nitride layer at the source side to that at the drain side), and because the (2) electric field intensity of the metal gate became weak. In regard to the fabrication process flow, the parameter  $D_{BCAT}$  can be controlled by adjusting the amount of the metal gate etch-back. If the metal gate etch-back or cleaning process is stable in the process deviation, it may provide a useful option for achieving the higher SS, on-/off- current, and lower DIBL [21].

However, the purpose of the stacking nitride insulator layer on top of the metal gate is to reduce gate-induced drain leakage (GIDL) by isolating the metal gate and drain and reducing the metal gate-drain overlapped region. Thus, the D<sub>BCAT</sub> must be adjusted and limited to the appropriate level in order to meet the device specifications, such as the GIDL. Increasing the D<sub>junction</sub> resulted in lower V<sub>th</sub>, worse DIBL, and non-steeper SS. This is due to the shortening of the effective channel length with the increasing  $D_{\text{junction}}$  [22]. Varying the D<sub>iunction</sub> might yield a better device performance without changing the physical dimensions of the device; thus, it has the advantages of avoiding any undesired defections caused by the process/structure. It was shown that increasing the  $D_{junction}$  (i.e., -33% to +33%of baseline) resulted in remarkable decrease in the V<sub>th</sub> (from 0.664 V to 0.641 V; nominal  $V_{th}$  = 0.656 V) and increase in the on-/off-current ratio (from 2.4  $\times$  10<sup>8</sup> to 5.0  $\times$  10<sup>11</sup>; nominal on/off-current ratio =  $3.4 \times 10^{10}$ ). However, as the DIBL increases from 21.0 mV/V to 28.2 mV/V, caused by the SCE, the control of the D<sub>iunction</sub> must be carefully considered by compensating for the SCE (i.e., silicon-on-insulator or junction engineering, such as pocket implanting, etc.). A narrower fin width ( $W_{fin}$ ) resulted in a lower  $V_{th}$ , non-steeper SS, and higher on-/off-current ratio. The channel region becomes fully depleted with the narrower fin width, so that the gate controllability of the channel region is enhanced [23]. A narrower W<sub>fin</sub> would require more advanced technologies in the fabrication process (i.e., smaller-scale lithography or etching processes), as the lateral size of the active Si becomes smaller. Similar to those of the ARgate, the risks of the leaning/bending of the active silicon substrate must be considered, since the aspect ratio of the height of the active silicon to W<sub>fin</sub>

increases as the  $W_{fin}$  decreases. Compared to the other parametric variations, the parameter,  $R_{fillet}$ , resulted in the least significant variation (i.e., <5%) in the device performance. This is because the saddle fin width was less than the fin height ( $W_{fin} = 17 \text{ nm}$ ,  $H_{fin} = 48 \text{ nm}$ ). Otherwise, the  $R_{fillet}$  would have affected the device performance. If the corner of the saddle fin becomes more rounded, the device reliability will be less degraded due to the less concentrated electric field at the corner [24,25].



**Figure 3.** Performance metrics of BCAT: (**a**) threshold voltage (V<sub>th</sub>), (**b**) subthreshold swing (SS), (**c**) on-/off-current ratio, and (**d**) drain-induced barrier lowering (DIBL) for given parametric variations.

### 4. Conclusions

The buried-channel-array transistor (BCAT) with a 20 nm-long gate length was simulated with the Sentaurus TCAD tool, and then the impacts of the structure variations on its device characteristics were investigated. For the given baseline device structure, the structure parameters including AR<sub>gate</sub>, D<sub>BCAT</sub>, D<sub>junction</sub>, W<sub>fin</sub>, and R<sub>fillet</sub> were adjusted in order to quantitatively observe the variations in the input/output characteristics and key device performance metrics (i.e., V<sub>th</sub>, SS, on-/off-current ratio, and DIBL). When the D<sub>junction</sub> was altered (i.e.,  $\pm 33\%$ ), the V<sub>th</sub> variation (between 0.664 V and 0.641 V; nominal V<sub>th</sub> = 0.656 V) and DIBL (from 21.0 mV/V to 28.2 mV/V; nominal DIBL = 23.6 mV/V) were most significantly degraded/affected. When the AR<sub>gate</sub> was increased by +17% (vs. the baseline), the on-/off-current ratio of the baseline was  $3.4 \times 10^{10}$ ). Among the other structural parameters, it turned out that R<sub>fillet</sub> minimally affected the device performance (i.e., <5%). Those structural variations, in the end, affected the gate-to-channel capacitances, effective channel length, and depletion regions of the BCAT.

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