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**Abstract:** This study investigates the operational characteristics of AlGaN/GaN high-electronmobility transistors (HEMTs) by employing various passivation materials with different dielectric constants and passivation structures. To ensure the simulation reliability, the parameters were calibrated based on the measured data from the fabricated basic  $Si<sub>3</sub>N<sub>4</sub>$  passivation structure of the HEMT. The Si<sub>3</sub>N<sub>4</sub> passivation material was replaced with high-k materials, such as  $Al_2O_3$  and HfO<sub>2</sub>, to improve the breakdown voltage. The Al<sub>2</sub>O<sub>3</sub> and HfO<sub>2</sub> passivation structures achieved breakdown voltage improvements of 6.62% and 17.45%, respectively, compared to the basic  $Si<sub>3</sub>N<sub>4</sub>$ passivation structure. However, the increased parasitic capacitances reduced the cut-off frequency. To mitigate this reduction, the operational characteristics of hybrid and partial passivation structures were analyzed. Compared with the HfO<sub>2</sub> passivation structure, the HfO<sub>2</sub> partial passivation structure exhibited a 7.6% reduction in breakdown voltage but a substantial 82.76% increase in cut-off frequency. In addition, the HfO<sub>2</sub> partial passivation structure exhibited the highest Johnson's figure of merit. Consequently, considering the trade-off relationship between breakdown voltage and frequency characteristics, the HfO<sub>2</sub> partial passivation structure emerged as a promising candidate for high-power and high-frequency AlGaN/GaN HEMT applications.

**Keywords:** gallium nitride; high-electron-mobility transistor; passivation; dielectric material; breakdown voltage

### **1. Introduction**

AlGaN/GaN high-electron-mobility transistors (HEMTs) are increasingly being adopted for high-power applications due to their superior material properties, such as a wide energy bandgap (3.4 eV) and a high critical electric field (3.39 MV/cm). GaN exhibits a higher electron saturation velocity and current density than conventional silicon and gallium arsenide [\[1](#page-12-0)[–3\]](#page-12-1). AlGaN/GaN heterostructure HEMTs exhibit exceptional performance due to the formation of a two-dimensional electron gas (2-DEG) at the surface, which arises from spontaneous and piezoelectric polarization effects [\[4,](#page-12-2)[5\]](#page-12-3). Consequently, these HEMTs are extensively employed in power electronics and devices that operate under high-power and high-frequency conditions. To optimize these outstanding characteristics, we developed various dielectric passivation structures that enhance the breakdown voltage ( $V_{BD}$ ) and cut-off frequency  $(f_T)$  simultaneously. After conducting the simulation for each structure, Johnson's figure of merit (JFOM), which can be expressed as  $V_{BD} \times f_T$ , was used to evaluate the operational characteristics [\[6](#page-12-4)[,7\]](#page-12-5).



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High-k materials are commonly used in the passivation layer because of their ad-High-k materials are commonly used in the passivation layer because of their advantages. Due to its higher dielectric constant than conventional materials such as  $SiO<sub>2</sub>$ vantages. But to its higher dielectric constant than conventional materials such as  $502$  and  $Si<sub>3</sub>N<sub>4</sub>$ , a high-k material functions as a thicker dielectric layer without physically increasing its thickness. It effectively reduces leakage current under both off-state and on-state conditions [\[8–](#page-12-6)[10\]](#page-12-7). In addition, high-k materials contribute to the electric field redis-state conditions [8–10]. In addition, high-k materials contribute to the electric field redis-tribution, which improves the breakdown voltage of HEMT devices [\[11](#page-12-8)[,12\]](#page-12-9). However, the large dielectric constant of these materials also increases parasitic capacitances, such as the gate-to-source capacitance ( $G<sub>g</sub>$ ), which capacitance (CGD), which can degate-to-source capacitance  $(C_{\text{GS}})$  and gate-to-drain capacitance  $(C_{\text{GD}})$ , which can degrade the frequency characteristics [\[13\]](#page-12-10). Therefore, it is crucial to balance the trade-off between V<sub>BD</sub> and frequency characteristics when selecting high-k materials for the passivation layer<br>since MGaN (GaN HEMTs [14,15] of AlGaN/GaN HEMTs [\[14,](#page-12-11)[15\]](#page-12-12). creasing its thickness. It is the contractively reduced current under the property of the state and ontribution, which is the breakdown voltage of HEMT devices the Breakdown voltage of HTMT devices  $\frac{1}{11}$ 

In this study, we simulated and analyzed three distinct passivation structures: entire In this study, we simulated and analyzed three distinct passivation structures: entire passivation (EP), hybrid passivation (HP), and partial passivation (PP) structures, using passivation (EP), hybrid passivation (HP), and partial passivation (PP) structures, using different dielectric materials such as  $Si_3N_4$ ,  $Al_2O_3$ , and  $HfO_2$  to improve the V<sub>BD</sub> with  $m$  in frequency calculated for a nalyzed for an local for an local for an local for all  $m$ minimal degradation in frequency characteristics. The JFOM was calculated and analyzed for seven different structures in terms of the trade-off relationship between  $V_{BD}$  and  $f_T$ . First, the  $Al_2O_3$  and HfO<sub>2</sub> EP structures were modeled by substituting the passivation  $Al_2O_3$ material of the basic  $Si_3N_4$  passivation structure. We confirmed that the EP structure with  $_{\text{H}}$ HfO<sup>2</sup> passivation, which has the highest dielectric constant among the dielectric materials, exhibited the highest V<sub>BD</sub> because it effectively redistributed the electric field when a highest V<sub>BD</sub> because it effectively redistributed the electric field when a high  $d$  drain voltage ( $V_{DS}$ ) was applied. Conversely, when a high-k material was applied as the passivation layer, the parasitic capacitances also increased, leading to the degradation of passivation layer, the parasitic capacitances also increased, leading to the degradation of  $f_T$  [\[16–](#page-12-13)[18\]](#page-12-14). To minimize the degradation of  $f_T$  caused by the use of a high-k material in the passivation layer, we suggest the use of HP and PP structures to improve the frequency passivation layer, we suggest the use of HP and PP structures to improve the frequency characteristics. The AlGaN/GaN HEMT with a properly designed dielectric passivation<br>dielectric passivation structure with high  $V_{BD}$  and  $f_T$  is expected to be a good candidate for high-power and high-frequency applications, such as GaN monolithic microwave integrated circuit power amplifiers for military radars and GaN radio frequency (RF) electronic devices for  $5<sup>th</sup>$  amplifiers for military radars and GaN radio frequency (RF) electronic devices for  $5<sup>th</sup>$ generation mobile telecommunication and autonomous driving. discretion (ET), hy ond passivation (TT), and partial passivation (TT) structures, as in  $t_{\text{tot}}$  several different structures in terms of the trade-on relationship between  $v_{\text{BD}}$  and  $r_{\text{1}}$ for the degradation degradation of the degradation of  $\frac{1}{2}$  and  $\frac{1}{2}$  material in the use of a high-k material in the use of a high-k material in the use of  $\frac{1}{2}$  material in the use of a high-k material in the characteristics. The AlGaN/GaN HEMT with a properly designed distribution of interesting passivation of the AlG ation mobile telecommunication and automobile telecommunication.

## **2. Materials and Methods** 2. Materials and Methods

<span id="page-1-0"></span>A 0.16  $\mu$ m T-gate AlGaN/GaN HEMT was fabricated, and a cross-sectional view of the unit device is shown in Figure [1a](#page-1-0). Figure [1b](#page-1-0) shows a magnified image of the gate electrode, featuring a 0.16 µm gate foot opening in the 1st passivation layer, which is covered on top by a 2nd passivation layer.





Figure [2](#page-2-0) illustrates a cross-sectional view of a 0.16  $\mu$ m T-gate AlGaN/GaN HEMT, which was used for technology computer-aided design (TCAD) modeling. In this figure, S

is the source electrode, S-FP is the source-connected field plate, G is the gate electrode, and D is the drain electrode. The specific geometric parameters of the modeled device are listed in Table [1.](#page-2-1)  $\frac{1}{2}$ 

<span id="page-2-0"></span>

Figure 2. An illustration of 0.16  $\mu$ m gate foot length of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP HEMT used for TCAD modeling. S-FP stands for source-connected field plate. G, D, and S stand for gate, drain, and respectively. source, respectively.

<b>Parameters</b>	Value $(\mu m)$			
1 L <sub>Source-Drain</sub>	4.585			
2 L <sub>Gate-Head-Top</sub>	0.26			
(3) L <sub>Gate-Head-Bottom</sub>	0.71			
$\textcircled{4}$ L <sub>Gate-Foot</sub>	0.16			
$\circled{5}$ L <sub>Gate-Drain</sub>	3.175			
6 L <sub>Gate-Source</sub>	0.7			
SiC-4H substrate	5			
Nucleation layer	0.02			
GaN buffer	1.04			
AlN insertion layer	0.001			
AlGaN barrier	0.018			
1st passivation	0.02			
2nd passivation	0.25			

<span id="page-2-1"></span>Table 1. Specific geometric parameters of the 0.16 µm gate foot length of the basic Si3N4 EP HEMT. **Table 1.** Specific geometric parameters of the 0.16  $\mu$ m gate foot length of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP HEMT.

The AlGaN/GaN HEMT was grown on top of a 4-inch SiC-4H substrate using metal– organic chemical vapor deposition. The epitaxial layers were sequentially stacked and grown as follows: a 20 nm thick nucleation layer, a  $1.04 \mu m$  thick Fe-doped GaN buffer layer, a 1 nm thick AlN insertion layer, and an 18 nm thick AlGaN barrier layer with 28% Al composition. The Ti/Au/Ni/Au alloyed ohmic contacts for the source and drain electrodes were formed by rapid thermal annealing at 775 ◦C for 30 s. Device isolation was achieved via  $P^+$  ion implantation. Subsequently, a 20 nm thick  $Si<sub>3</sub>N<sub>4</sub>$  layer was deposited on the AlGaN barrier layer using plasma-enhanced chemical vapor deposition (PECVD). The first metal interconnection with the source and drain electrodes was established by Ti/Au evaporation after the etching of the 1st  $Si<sub>3</sub>N<sub>4</sub>$  passivation layer. A planar gate was then created using single-layer electron beam lithography. A gate foot opening of 0.16 µm was achieved by exposing a polymethyl methacrylate resist to an electron beam, followed by the removal of the 1st  $Si<sub>3</sub>N<sub>4</sub>$  passivation layer beneath the gate foot opening pattern through dry etching using inductively coupled plasma. The planar gate was defined using a Ni/Au metal stack deposited via electron–beam evaporation and subsequent lift-off processes. After defining the gate shape, a 250 nm thick 2nd  $Si<sub>3</sub>N<sub>4</sub>$  passivation layer was deposited for device passivation using PECVD. The source-connected field plate (S-FP) was formed using

a Ti/Au metal lift-off process. Finally, wafer thinning and backside via-hole processes were conducted [\[19\]](#page-12-15).

To accurately predict the operational characteristics of a device, it is crucial to apply appropriate simulation parameters, such as electrical and thermal parameters, for each epitaxial layer. This meticulous approach ensures reliable and consistent simulation data. Consequently, the simulation parameters were meticulously calibrated to closely align with the actual device operating characteristics. For example, to mitigate the electron punch-through effect and reduce the substrate leakage current, iron (Fe) acceptor trap doping was leveraged in the GaN buffer layer to enhance the  $V_{BD}$  [\[20\]](#page-12-16). In this simulation, a Gaussian acceptor doping profile was employed, with an acceptor doping concentration of  $8.813 \times 10^{14}/\text{cm}^3$  at the AlGaN/GaN interface region and a peak trap concentration of  $10^{18}/\text{cm}^3$  [\[21\]](#page-13-0). In addition, a Selberherr impact ionization model was applied to simulate the  $V_{BD}$ . Other simulation parameters such as electron mobility and heat models were accurately controlled to obtain reliable simulation results. The specific simulation parameters applied to the GaN and AlGaN layers are summarized in Table [2](#page-3-0) [\[22\]](#page-13-1).

<span id="page-3-0"></span>**Table 2.** Material parameters for the simulation at a room temperature.



After determining the appropriate simulation parameters, simulations were conducted to analyze the direct current (DC) and RF characteristics. The transconductance equation can be expressed as follows:

$$
g_{\rm m} = \frac{\partial I_{\rm DS}}{\partial V_{\rm GS}},\tag{1}
$$

where  $g_{m}$ , I<sub>DS</sub>, and V<sub>GS</sub> denote the transconductance, drain current, and gate voltage, respectively. The electric displacement was explained by Equation (2), as follows:

$$
D = \varepsilon E, \tag{2}
$$

where  $D$ ,  $\varepsilon$ , and E denote the electrical displacement, dielectric constant of the material, and electric field, respectively. Before evaluating the frequency characteristics of each structure, the relationship between the parasitic capacitances, such as  $C_{\text{GS}}$  and  $C_{\text{GD}}$ , and the frequency characteristics was investigated as follows:

$$
C = \frac{\varepsilon A}{d},\tag{3}
$$

where A and d denote the overlapped area between two electrodes and the distance between the electrodes, respectively.

Next,  $f_T$  can be determined using Equation (4), as follows:

$$
f_{\rm T} = \frac{g_{\rm m}}{2\pi (C_{\rm GS} + C_{\rm GD})} \approx \frac{g_{\rm m}}{2\pi C_{\rm GS}},\tag{4}
$$

where  $C_{\text{GS}}$  and  $C_{\text{GD}}$  denote the gate-to-source capacitance and gate-to-drain capacitance, respectively. As described in Equation (4),  $C_{\text{GS}}$  and  $C_{\text{GD}}$  have an inverse relationship with

 $\rm f_T$ , which makes it crucial to minimize parasitic capacitances to maximize the frequency characteristics [\[23\]](#page-13-2). Therefore, we propose various dielectric passivation structures using materials with different dielectric constants, such as  $Si<sub>3</sub>N<sub>4</sub>$ , Al<sub>2</sub>O<sub>3</sub>, and HfO<sub>2</sub>, to analyze the RF characteristics related to capacitances. The specific material characteristics of these materials are summarized in Table [3](#page-4-0) below [\[24](#page-13-3)[,25\]](#page-13-4).  $\rm r_T$ , which makes it crucial to minimize parasitic capacitances to maximize the frequency

respectively. As described in Equation (4), Cୋୗ and Cୋୈ have an inverse relationship

<span id="page-4-0"></span>**Table 3.** Material characteristics of  $Si<sub>3</sub>N<sub>4</sub>$ ,  $Al<sub>2</sub>O<sub>3</sub>$ , and  $HfO<sub>2</sub>$ .

Parameters	Units	Si <sub>3</sub> N <sub>4</sub>	$Al_2O_3$	HfO <sub>2</sub>
Dielectric constant	-	~27.5	$\sim$ 9	~25
Bandgap energy	eV	5.3		

#### **3. Results**

#### *3.1. Matching Simulated and Measured Data for the Basic Si3N<sup>4</sup> Entire Passivation Structure*  $3.$  Nesults  $\sim$  Measured Data for the Basic Si $3.1$  Entire Passivation Structure Passiv  $T_{\text{tot}}$  and simulation accuracy, and comparative accuracy, a conduction between  $\frac{1}{2}$

To validate the simulation accuracy, a comparative analysis was conducted between the simulated and measured drain current–gate voltage  $(I_{DS} - V_{GS})$  transfer characteristics of the fabricated  $Si<sub>3</sub>N<sub>4</sub>$  EP structure device. The measured and simulated data exhibited close agreement in terms of  $I_{DS}$  at  $V_{GS} = 0$  V ( $I_{dss}$ ), maximum transconductance ( $G_m$ ), and  $G_{L}$ threshold voltage ( $V_{th}$ ). Figure [3a](#page-4-1) compares the measured and simulated  $I_{DS}$ - $V_{GS}$  transfer characteristics. The measured and simulated  $I_{dss}$  values were 817.10 and 811.99 mA/mm, respectively. Similarly, the measured and simulated maximum transconductance values respectively. Similarly, the measured and simulated maximum transconductance values were 400.39 and 397.65 mS/mm, respectively. Furthermore, the measured V<sub>th</sub> was  $-3.1$  V, and the simulated value was  $-3$  V. These results confirm a close match between the measured and simulated data for  $I_{dss}$ ,  $G_m$ , and  $V_{th}$  with error rates of 0.6%, 0.7%, and  $3.2$ 3.2%, respectively. spectively.

<span id="page-4-1"></span>

Figure 3. Measured and simulated results of the basic Si<sub>3</sub>N<sub>4</sub> EP HEMT: (a) drain current–gate voltage  $(I_{DS}-V_{GS})$  transfer characteristics at drain voltage (V<sub>DS</sub>) = 10 V; (**b**) cut-off frequency (f<sub>T</sub>) at V<sub>DS</sub> = 10 V and gate voltage  $(V_{GS}) = -2 V$ .

The measured and simulated  $f_T$  values of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP structure are shown in Figure [3b](#page-4-1). The RF characteristics were evaluated under  $V_{DS}$  = 10 V and  $V_{GS}$  = −2 V conditions for both measurement and simulation. More specifically,  $f_T$  was defined as the intersection of the extension line at the current gain point  $(H_{21})$  with the *x*-axis with a slope of  $-20$  dB/decade [\[26\]](#page-13-5). The measured and simulated f<sub>T</sub> values were 29.26 and 29.51 GHz, respectively, demonstrating excellent agreement with the minimal error rate of 0.9%.

### *3.2. Comparative Analysis of Entire Passivation Structures Based on Dielectric Materials*

To accommodate high-power applications, the passivation layer of the  $Si<sub>3</sub>N<sub>4</sub>$  EP structure was replaced with a high-k material. Two distinct dielectric materials were <span id="page-5-0"></span>modeled  $(Al_2O_3$  and  $HfO_2$ ) for the EP structure, as shown in Figure [4.](#page-5-0) All structural parameters except for the passivation material remained unchanged during the simulation. modeled  $\left(\text{Al}^2\text{O}^3\right)$  and  $\left(\text{Al}^2\text{O}^3\right)$  for the ET structure, as shown in Figure 4. All



Figure 4. Illustrations of EP structures: (a) with  $Al_2O_3$ ; (b) with HfO<sub>2</sub>. S-FP stands for sourcenected field plate. G, D, and S stand for gate, drain, and source, respectively. connected field plate. G, D, and S stand for gate, drain, and source, respectively. rigule 4. Inustrations of El structures. (a) with  $\text{Al}_2\text{O}_3$ , (b) with  $\text{Al}_2\text{O}_2$ .

## 3.2.1. Simulation Results of the DC Characteristics 3.2.1. Simulation Results of the DC Characteristics 3.2.1. Simulation Results of the DC Characteristics

The DC characteristics of the  $\text{Al}_2\text{O}_3$  and  $\text{HfO}_2$  EP structures were compared with those of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP structure. Figure [5](#page-5-1) shows the I<sub>DS</sub>-V<sub>GS</sub> transfer characteristics of the three structures at  $V_{DS}$  = 10 V. No significant variations in the  $I_{DS}$  were observed, and the V<sub>th</sub> remained constant at  $-3.0$  V.

<span id="page-5-1"></span>

Figure 5. Simulation results of  $I_{DS}$ -V<sub>GS</sub> transfer characteristics for the three EP structures at V<sub>DS</sub> = 10 V.

for the three passivation structures. Compared with the  $Si<sub>3</sub>N<sub>4</sub>$  EP structure, the  $Al<sub>2</sub>O<sub>3</sub>$ and  $HfO<sub>2</sub>$  EP structures exhibited more efficient electric field dispersion, resulting in a lower maximum electric field in the channel layer due to their high dielectric constant. As the maximum electric field increased, the impact ionization that caused the generation of electron-hole pairs was enhanced; therefore, electric field redistribution effectively improved  $V_{BD}$  [\[27\]](#page-13-6). The dielectric constant of  $Al_2O_3$  is lower than that of HfO<sub>2</sub>, resulting in a relatively lower V<sub>BD</sub> [\[28\]](#page-13-7). The V<sub>BD</sub> characteristics were simulated under a V<sub>GS</sub> =  $-7$  V pinch-off condition to ensure a completely off device state. We defined  $V_{BD}$  as the  $V_{DS}$ Figure [6a](#page-6-0) shows the electric field distributions within the channel layer at  $V_{DS} = 500$  V when the  $I_{DS}$  exceeded 1 mA/mm after completely turning off the device by applying a voltage of  $-7$  V across the gate. As shown in Figure [6b](#page-6-0), the Si<sub>3</sub>N<sub>4</sub>, Al<sub>2</sub>O<sub>3</sub>, and HfO<sub>2</sub> EP structures exhibited V<sub>BD</sub> values of 519.97, 554.39, and 610.70 V, respectively. The V<sub>BD</sub> of the  $\text{Al}_2\text{O}_3$  and HfO<sub>2</sub> EP structures were improved by 6.62% and 17.45%, respectively, compared with that of the  $Si<sub>3</sub>N<sub>4</sub>$  EP structure.



<span id="page-6-0"></span>the Si<sup>3</sup>N4 E<sub>P</sub> structure. The Si<sup>3</sup>N4 E<sub>P</sub> structure.

Figure 6. (a) Electric field distributions across the 2-DEG (2-dimensional electron gas) channel layer between the source and drain electrodes at  $V_{DS} = 500$  V and  $V_{GS} = -7$  V; (b) breakdown voltage  $(V_{BD})$  at pinch-off ( $V_{GS} = -7 V$ ).

# 3.2.2. Simulation Results of the RF Characteristics of the parasitic capacitance characteristics of the  $R$ F Characteristics of the  $\mathbb{R}$

<span id="page-6-1"></span>Figur[e 7](#page-6-1) shows the parasitic capacitance characteristics of the  $\rm{Si_3N_4}$  ,  $\rm{Al_2O_3}$  , and  $\rm{HfO_2}$ EP structures. As shown in Figur[e 7](#page-6-1)a,b, the C<sub>GS</sub> and C<sub>GD</sub> were determined at  $V_{DS} = 10 V$ and  $V_{GS} = -2 V$ . The HfO<sub>2</sub> EP structure exhibited the highest C<sub>GS</sub> and C<sub>GD</sub> values, which can be attributed to the dielectric constant of  $HfO_2$ , as described by Equation (3). Conversely, the  $Al_2O_3$  EP structure exhibited lower parasitic capacitance values than the  $HfO_2$ EP structure, due to its lower dielectric constant.



Figure 7. Simulated capacitance characteristics as a function of frequency for three different EP structures at  $V_{DS}$  = 10 V and  $V_{GS}$  = -2 V; (a) gate-to-source capacitance (C<sub>GS</sub>) and (b) gate-to-drain capacitance (CGD). capacitance (CGD). capacitance (CGD).

Figure [8](#page-7-0) shows the simulated  $f_T$  and  $V_{BD}$  values for the three EP structures.  $f_T$ simulations were conducted at  $V_{DS}$  = 10 V and  $V_{GS}$  = -2 V. According to Equation (4), the  $f_T$  values of the three EP structures were affected by transconductance  $(g_m)$  and  $C_{GS}$ . The  $Si<sub>3</sub>N<sub>4</sub>$ , Al<sub>2</sub>O<sub>3</sub>, and HfO<sub>2</sub> EP structures exhibited f<sub>T</sub> values of 29.51, 28.16, and 16.07 GHz, respectively. Compared with the  $Si<sub>3</sub>N<sub>4</sub>$  EP structure, the  $Al<sub>2</sub>O<sub>3</sub>$  and HfO<sub>2</sub> EP structures exhibited reductions of 4.57% and 45.54%, respectively.

<span id="page-7-0"></span>

Figure 8. Simulated  $f_T$  and  $V_{BD}$  for three different EP structures.

3.2.3. Simulation Results of the Hybrid Passivation Structure 3.2.3. Simulation Results of the Hybrid Passivation Structure 3.2.3. Simulation Results of the Hybrid Passivation Structure

<span id="page-7-1"></span>To address the trade-off between enhanced  $\rm V_{BD}$  values and degraded  $\rm f_T$  associated with the application of  $Al_2O_3$  and  $HfO_2$  to the EP structures, HP structures were proposed by employing  $Al_2O_3$  and  $\overline{H_2O_2}$  into the 1st passivation and  $Si_3N_4$  into the 2nd passivation. Figure [9](#page-7-1) shows the schematics of the HP structures with Al<sub>2</sub>O<sub>3</sub> and HfO<sub>2</sub>.



Figure 9. Illustrations of hybrid passivation (HP) structures: (a) with  $Al_2O_3$ ; (b) with  $HfO_2$ .

Figure [10a](#page-8-0) shows the electric field distributions for three different structures. The maximum electric fields of the  $Al_2O_3$  and  $HfO_2$  HP structures were lower than those of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP structure. The dielectric constant of  $Al<sub>2</sub>O<sub>3</sub>$  is lower than that of HfO<sub>2</sub>, resulting in a relatively lower  $V_{BD}$ . Specifically, the  $V_{BD}$  values of the  $Al_2O_3$  and  $HfO_2$ HP structures were 546.39 and 572.87 V, respectively, as shown in Figure 10b. How[ever](#page-8-0), compared with the EP structure, the HP structure exhibited a reduced  $V_{BD}$  because of the use of a high-k material only for the 1st passivation.

Figure [11](#page-8-1) shows the parasitic capacitance characteristics of the different 1st passivation materials. Given that  $L_{\text{Gate-Source}}$  was shorter than  $L_{\text{Gate-Drain}}$ ,  $C_{\text{GS}}$  exhibited a larger value than  $C_{GD}$ , indicating that the capacitance was affected by the distance between the electrodes [\[29\]](#page-13-8). Figure [11](#page-8-1) shows that the HfO<sub>2</sub> HP structure exhibited the highest  $C_{\text{GS}}$ and  $C_{GD}$  values. Conversely, the  $Al_2O_3$  HP structure exhibited lower parasitic capacitance values than the  $HfO<sub>2</sub>$  HP structure, which was due to the relatively low dielectric constant of  $Al_2O_3$ .

Figure [12](#page-8-2) compares the simulated  $f<sub>T</sub>$  values for the three dielectric passivation structures. Simulations conducted at  $V_{DS} = 10$  V and  $V_{GS} = -2$  V revealed  $f_T$  values of 28.63 and 26.46 GHz for the  $Al_2O_3$  and  $HfO_2$  HP structures, respectively. Compared with the  $Si<sub>3</sub>N<sub>4</sub>$  EP structure, these values represent  $f<sub>T</sub>$  reductions of 2.98% and 10.34% for the HfO<sub>2</sub> and  $\text{Al}_2\text{O}_3$  HP structures, respectively. According to Equation (4), the decrease in  $f_T$  can be attributed to the increase in  $C_{\text{GS}}$ . Compared to the high-k EP structure, the HP structure

<span id="page-8-0"></span>

compensated for the decrease in RF characteristics by applying a high-k material only at the 1st passivation layer.

<span id="page-8-1"></span>Figure 10. Comparison of  $Si<sub>3</sub>N<sub>4</sub>$  EP, Al<sub>2</sub>O<sub>3</sub> HP, and HfO<sub>2</sub> HP structures: (a) electric field distributions across the 2-DEG channel layer between the source and drain electrodes at  $V_{DS} = 500$  V and  $V_{\text{GS}} = -7 \text{ V}$ ; (**b**)  $V_{\text{BD}}$  at pinch-off ( $V_{\text{GS}} = -7 \text{ V}$ ).



Figure 11. Simulated capacitance characteristics as a function of frequency for  $Si<sub>3</sub>N<sub>4</sub>$  EP, Al<sub>2</sub>O<sub>3</sub> HP, and HfO<sub>2</sub> HP structures at V<sub>DS</sub> = 10 V and V<sub>GS</sub> = −2 V; (**a**) C<sub>GS</sub> and (**b**) C<sub>GD</sub>.

<span id="page-8-2"></span>

**Figure 12.** Simulated  $f_T$  and  $V_{BD}$  values for the  $Si_3N_4$  EP,  $Al_2O_3$  HP, and HfO<sub>2</sub> HP structures.

## *3.3. Comparative Analysis of Partial Passivation Structures Based on Al2O<sup>3</sup> and HfO<sup>2</sup>* 3.3. Comparative Analysis of Partial Passivation Structures Based on Al2O3 and HfO<sup>2</sup>

Figure 12. Simulated fT and VBD values for the Si3N4 EP, Al2O3 HP, and HfO2 HP structures.

To mitigate the degradation of the RF characteristics observed in the HP structure To mitigate the degradation of the RF characteristics observed in the HP structure while preserving the benefits of high-k materials, a PP structure was introduced. By implementing the HP structure, the RF characteristics were improved compared to the high-k EP structure. However, the  $f_T$  of the HP structure was lower than that of the basic  $Si<sub>3</sub>N<sub>4</sub>$  EP structure. To minimize the degradation of the RF characteristics, we applied the PP structure with a high-k material only for the 1st passivation layer at the drain-gate region. Figure [13](#page-9-0) shows schematic diagrams of the PP structure with  $\text{Al}_2\text{O}_3$  and HfO<sub>2</sub>.

<span id="page-9-0"></span>

**Figure 13.** Illustrations of partial passivation (PP) structures: (a) with  $Al_2O_3$ ; (b) with  $HfO_2$ .

3.3.1. Simulation Results of the DC Characteristics 3.3.1. Simulation Results of the DC Characteristics

The I<sub>DS</sub>-V<sub>GS</sub> transfer characteristics, electric field distribution, and V<sub>BD</sub> characteristics of the  $\text{Al}_2\text{O}_3$  and HfO<sub>2</sub> PP structures were simulated. As shown in Figur[e 14](#page-9-1), the I<sub>DS</sub>,  $g_m$ , and  $V_{th}$  values remained unaffected by variations in the material of the 1st passivation layer at the drain–gate region. layer at the drain–gate region.

<span id="page-9-1"></span>

Figure 14. Simulation results of I<sub>DS</sub>-V<sub>GS</sub> transfer characteristics in the Si<sub>3</sub>N<sub>4</sub> EP, Al<sub>2</sub>O<sub>3</sub> PP, and HfO<sub>2</sub> PP structures. PP structures.

Figure [15a](#page-10-0) shows that the maximum electric field for the  $HfO<sub>2</sub>$  PP structure, which exhibited the highest dielectric constant, decreased and was dispersed in the drain–gate exhibited the highest dielectric constant, decreased and was dispersed in the drain–gate region. Conversely, the lower dielectric constant of  $\text{Al}_2\text{O}_3$  in the  $\text{Al}_2\text{O}_3$  PP structure resulted in less pronounced electric field dispersio[n. F](#page-10-0)igure 15b shows that the V<sub>BD</sub> of the HfO<sub>2</sub> PP structure exhibited the highest V<sub>BD</sub> value of 564.27 V, while the  $Si<sub>3</sub>N<sub>4</sub>$  EP and Al<sub>2</sub>O<sub>3</sub> PP structures exhibited comparable values of 519.97 and 532.08 V, respectively. Notably, the use of a high-k material as the 1st passivation layer at the drain–gate region, where the electric field peak occurs, resulted in a slight decrease in  $\rm{V_{BD}}$  for the PP structure compared to the HP structure. to the HP structure.



<span id="page-10-0"></span>to the HP structure. The HP structure is the HP structure.

Figure 15. Comparison of Si<sub>3</sub>N<sub>4</sub> EP, Al<sub>2</sub>O<sub>3</sub> PP, and HfO<sub>2</sub> PP structures: (a) electric field distributions across the 2-DEG channel layer between the source and drain electrodes at  $V_{DS} = 500$  V and  $V_{GS} = -7 V$ ; (**b**)  $V_{BD}$  at pinch-off ( $V_{GS} = -7 V$ ).

### <span id="page-10-1"></span>3.3.2. Simulation Results of the RF Characteristics 3.3.2. Simulation Results of the RF Characteristics

Figure [16](#page-10-1) shows the parasitic capacitance characteristics of the  $Si<sub>3</sub>N<sub>4</sub>$  EP, HfO<sub>2</sub>, and  $Al_2O_3$  PP structures. Given that all three structures employed  $Si_3N_4$  as a passivation layer at the source–gate region, the  $C_{GS}$  remained consistent, as shown in Figure [16a](#page-10-1). However, Figure [16b](#page-10-1) shows that the HfO<sub>2</sub> PP structure exhibits the highest  $C_{GD}$  value.



Figure 16. Simulated capacitance characteristics as a function of frequency for Si<sub>3</sub>N<sub>4</sub> EP, Al<sub>2</sub>O<sub>3</sub> PP, and HfO<sub>2</sub> PP structures at V<sub>DS</sub> = 10 V and V<sub>GS</sub> = −2 V; (**a**) C<sub>GS</sub> and (**b**) C<sub>GD</sub>.

Figure [17 s](#page-11-0)hows the simulated  $f_T$  values for different dielectric passivation structures at V<sub>DS</sub> = 10 V and V<sub>GS</sub> = −2 V. Notably, the f<sub>T</sub> values of the Si<sub>3</sub>N<sub>4</sub> EP, Al<sub>2</sub>O<sub>3</sub> HP, and HfO<sub>2</sub> HP structures exhibited minimal variations (29.51, 29.44, and 29.37 GHz, respectively). HP structures exhibited minimal variations (29.51, 29.44, and 29.37 GHz, respectively). Equation (4) indicates that f<sub>T</sub> is primarily influenced by C<sub>GS</sub>, and a negligible change in  $C_{\text{GS}}$  results in the observed  $f_{\text{T}}$  consistency. These results highlight the effectiveness of the structure in mitigating the degradation of the RF characteristics. PP structure in mitigating the degradation of the RF characteristics.

<span id="page-11-0"></span>

structure in mitigating the degradation of the degradation of the RF characteristics. The RF characteristics.

Figure 17. Simulated  $f_T$  and  $V_{BD}$  for  $Si_3N_4$  EP,  $Al_2O_3$  PP, and  $HfO_2$  PP structures.

### 4. Discussion **4. Discussion**

This study simulates and analyzes the DC and RF characteristics of various dielectric This study simulates and analyzes the DC and RF characteristics of various dielectric passivation structures. Table 4 summarizes the DC and RF characteristics, including passivation structures. Table [4](#page-11-1) summarizes the DC and RF characteristics, including JFOM, for seven different dielectric passivation structures of the AlGaN/GaN HEMT. Among the Si<sub>3</sub>N<sub>4</sub>, Al<sub>2</sub>O<sub>3</sub>, and HfO<sub>2</sub> EP structures, the HfO<sub>2</sub> EP structure exhibited the highest  $V_{BD}$ . However, the high-k passivation layer inevitably entailed a decrease in  $f_T$  due to  $V_{BD}$ . parasitic capacitance. To minimize the degradation of  $r_1$ , HP and PP structures were applied. The JFOM was calculated to analyze the trade-off relationship between  $V_{BD}$  and  $f_T$ . The basic  $Si_3N_4$  EP structure has a JFOM of 15.34 THz-V. The JFOMs with three different  $A<sub>1</sub>$ . The basic  $S<sub>3</sub>$  A<sub>4</sub> EP structures were not significantly different from the Si<sub>3</sub>N<sub>4</sub> EP structure. Algostration structures were not significantly different from the Si3N4 EP structure. However, the proposed HfO<sub>2</sub> PP structure exhibited the highest JFOM of 16.75 THz-V with  $\epsilon$ enhanced  $\rm V_{BD}$  while maintaining  $\rm f_{T.}$ parasitic capacitance. To minimize the degradation of  $f<sub>T</sub>$ , HP and PP structures were

<b>Parameters</b>	Units	Si <sub>3</sub> N <sub>4</sub>		$Al_2O_3$			HfO <sub>2</sub>	
Structure type	$\overline{\phantom{0}}$	EP	EP	ΗP	PP	EP	ΗP	PP
Peak electric field Breakdown voltage (V <sub>BD</sub> )	MV/cm	5.16 519.97	4.98 554.39	5.09 546.63	5.13 532.08	3.83 610.70	4.22 572.87	4.69 564.27
Cut-off frequency $(f_T)$	GHz	29.51	28.16	28.63	29.44	16.07	26.46	29.37
Johnson's figure-of-merit (JFOM)	THz-V	15.34	15.63	15.65	15.66	9.81	15.16	16.75

<span id="page-11-1"></span>**Table 4.** A summary of DC and RF characteristics of various dielectric passivation structures of HEMT.

### **5. Conclusions**

This study investigates the DC and RF characteristics of AlGaN/GaN HEMTs using various passivation material configurations via TCAD simulation. The simulation parameters were obtained by matching the simulation data with the measured data of a fabricated basic  $Si<sub>3</sub>N<sub>4</sub>$  EP structure of HEMT to ensure the reliability of the simulation results. The JFOM was calculated to assess the operational characteristics of each proposed dielectric passivation structure considering the trade-off between the breakdown voltage and cut-off frequency. Consequently, based on the highest calculated JFOM among the investigated structures, the HfO<sub>2</sub> PP structure was proposed as the optimal dielectric passivation structure for achieving superior breakdown voltage and frequency characteristics. This structure shows promise for high-power and high-frequency AlGaN/GaN HEMT applications.

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