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# Control Design of a Swarm of Intelligent Robots: A Closed-Form $H_2$ Nonlinear Control Approach

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**Abstract:** A closed-form  $H_2$  approach of a nonlinear trajectory tracking design and practical implementation of a swarm of wheeled mobile robots (WMRs) is presented in this paper. For the nonlinear trajectory tracking problem of a swarm of WMRs, the design purpose is to point out a closed-form  $H_2$  nonlinear control method that analytically fulfills the  $H_2$  control performance index. The key and primary contribution of this research is a closed-form solution with a simple control structure for the trajectory tracking design of a swarm of WMRs is an absolute achievement and practical implementation. Generally, it is challenging to solve and find out the closed-form solution for this nonlinear trajectory tracking problem of a swarm of WMRs. Fortunately, through a sequence of mathematical operations for the trajectory tracking error dynamics between the control of a swarm of WMRs and desired trajectories, this  $H_2$  trajectory tracking problem is equal to solve the nonlinear time-varying Riccati-like equation. Additionally, the closed-form solution of this nonlinear time-varying Riccati-like equation will be acquired with a straightforward form. Finally, for simulation-controlled performance of this  $H_2$  proposed method, two testing scenarios, circular and S type reference trajectories, were applied to performance verification.

**Keywords:** wheeled mobile robot (WMR); closed-form solution;  $H_2$  performance index

## 1. Introduction

Over the past few decades, scientific and technological progression and innovation have led to the universal application of wheeled mobile robots (WMRs) in daily life. These WMRs with more precise motion capability and powerful controllers are applied to inspection, security, and transportation, etc. According to the above, the accurate motion control of WMRs is more and more critical in the robotics industry, and many researchers have also stepped into this attractive topic in recent decades [1–6]. The real-time tracking control design of WMRs is always an essential subject for the controller with a straightforward implementation configuration. From the existing studies, it is still an open challenge question to find out how to effectively develop a control design of WMRs to precisely track the desired trajectories in robotics. The trajectory tracking design of the WMRs system has the capability of handling the tracking error between the actual path of the controlled WMRs and the desired trajectory to converge to approach zero under the effects of slippage, disturbances, and measurement noises. Therefore, the WMRs manipulating precisely is more and more critical in the trajectory tracking topic under external disturbance effects. By the survey of existing literature, many research results have worked on tracking control of WMRs, which focused on four categorizations: (1) sliding mode control [7–12]; (2) feedback linearization [13,14]; (3) backstepping [15–17]; and (4) neural networks and Fuzzy approaches [18–23]. Other designs have also stated the combination of the fuzzy theory or

sliding mode and neural network control, as presented in [19,21,23]. According to efficient experiences, it is challenging to implement low-cost microchips with computational consumption based on this kind of control algorithm methodology and very complicated theory structure.

Base on the above reasons, a progressive nonlinear control method, which provides an easy hardware implementation and a high-performance trajectory tracking for a swarm of WMRs, will be proposed in this recommended research. For achieving the objectives of easy hardware implementation and the lowest consumption in the computational calculation, a revolutionary nonlinear control method was inferred for the trajectory tracking problem of a swarm of WMRs. We try to solve a nonlinear and complex time-varying Riccati-like equation directly [24–26], and this is equal to solving an  $H_2$  closed-form solution for the trajectory tracking problem of a swarm of WMRs based on an optimal performance index. This  $H_2$  closed-form solution is also a key and primary contribution to this research. Considering the above statement, the subject of a nonlinear optimal control design must solve one nonlinear time-varying Riccati-like equation, which is very difficult to explain, and it is difficult to find out the solution. Fortunately, this closed-form solution can be derived from the appropriately selected state variable transformation and tracking error dynamics analysis in this research. According to this closed-form solution, one nonlinear optimal control method contains a straightforward implementation structure for the trajectory tracking problem of a swarm of WMRs that can be constructed. This paper presents the following sections: The introduction and literature review will be stated in Section 1; the mathematical model and tracking error dynamics of a swarm of WMRs will be described in Section 2; problem formulation and the optimal controller design objective with a closed-form solution for a swarm of WMRs trajectory tracking will be introduced in Section 3; and the simulation performance verifications of the trajectory tracking ability of a swarm of WMRs by the proposed method are demonstrated in Section 4. Finally, the conclusions are summarized in Section 5.

## 2. Mathematical Model and Tracking Error Dynamics

In this section, the mathematical model and the controlled description of a swarm of WMRs will be presented. According to this regular equation and the geometry relationship between a swarm of WMRs, a nonlinear tracking error dynamic equation for the controlled WMRs can be obtained.

### 2.1. Dynamic Model of a Swarm of WMRs

A typical model of the WMR, which contains two driving wheels and a passive self-adjusted supporting wheel, with the same radius express by  $r$  and separated by  $2R$ , is shown in Figure 1. The vector  $r$  exhibits the global coordinate frame  $\{O, X, Y\}$  of real-time location for WMR, which is the coordinate of the point  $C$  in the universal coordinate frame, and  $\theta$  is the orientation of the local frame  $\{C, X_c, Y_c\}$ . The distance between points  $r$  and  $C$  are denoted by  $d$ . Base on the above definitions, the generalized coordinate of the WMR can be represented as the equation below.

$$r = [ x_c \quad y_c \quad \theta ]^T \quad (1)$$

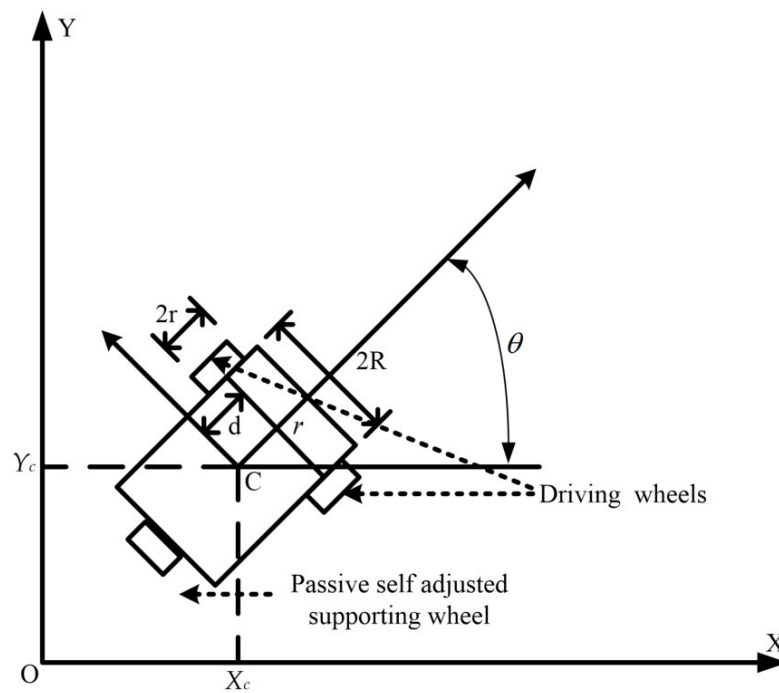


Figure 1. The wheeled mobile robot model.

A general WMR system, the robot can move in the direction of the axis of the driving wheels with pure rolling and non-slipping condition status, and the kinematic equation of the WMR under the constraint can be depicted as below [27]:

$$\dot{r} = \begin{bmatrix} \dot{x}_c \\ \dot{y}_c \\ \dot{\theta} \end{bmatrix} = \begin{bmatrix} \cos \theta & -d \sin \theta \\ \sin \theta & d \cos \theta \\ 0 & 1 \end{bmatrix} \begin{bmatrix} v_l \\ \omega \end{bmatrix} \quad (2)$$

where  $v_l$  and  $\omega$  are the velocity of linear and angular along the robot axis, respectively.

The above kinematic equation for WMR is used to infer the dynamics of WMR. In this research, the dynamic equation of a swarm of WMRs can be given as:

$$M_i(r_i)\ddot{r}_i + G_i(r_i, \dot{r}_i)\dot{r}_i + K_{gi}(r_i) = O_i(r_i)\tau_i \quad (3)$$

where  $M_i(r_i) \in \mathbb{R}^{3 \times 3}$  is a symmetric positive definite inertia matrix and  $G_i(r_i, \dot{r}_i) \in \mathbb{R}^{3 \times 3}$  is the centripetal and Coriolis matrix. Both  $M_i(r_i)$  and  $G_i(r_i, \dot{r}_i)$  satisfy the skew-symmetric property and  $K_{gi}(r_i) \in \mathbb{R}^{3 \times 3}$  is the gravitational vector. Since the WMR moves in the horizontal plane, the gravitational vector  $K_{gi}(r_i) \in \mathbb{R}^{3 \times 3}$  is zero,  $O_i(r_i) \in \mathbb{R}^{3 \times 2}$  is the input transformation matrix,  $\tau_i \in \mathbb{R}^{2 \times 1}$  is then applied to the torque vector,  $\dot{r}_i$  and  $\ddot{r}_i$ , and denotes velocity and acceleration vectors, and  $i$  is the number of WMRs. where:

$$M_i(r_i) = \begin{bmatrix} m_i & 0 & m_i d_i \sin \theta_i \\ 0 & m_i & -m_i d_i \cos \theta_i \\ m_i d_i \sin \theta_i & -m_i d_i \cos \theta_i & I \end{bmatrix}$$

$$G_i(r_i, \dot{r}_i) = \begin{bmatrix} 0 & 0 & m_i d_i \dot{\theta}_i \cos \theta_i \\ 0 & 0 & m_i d_i \dot{\theta}_i \sin \theta_i \\ 0 & 0 & 0 \end{bmatrix} O_i(r_i) = \frac{1}{r_i} \begin{bmatrix} \cos \theta_i & \sin \theta_i \\ \sin \theta_i & \cos \theta_i \\ R_i & -R_i \end{bmatrix}$$

$$\tau_i = \begin{bmatrix} \tau_{ri} \\ \tau_{li} \end{bmatrix}$$

where  $\tau_{ri}$  and  $\tau_{li}$  represent right and left wheel torques, respectively.

2.2. Problem Formulation

The desired trajectory tracking  $r_{ri}$  is supposed to exist in limited time functions of the position  $r_{ri} \in C^2$ , which is a twice continuously differentiable function. The velocity vector and acceleration vector  $r_{ri}$  can be expressed as  $\dot{r}_{ri}$  and  $\ddot{r}_{ri}$ , respectively. The definition of the tracking error between the desired trajectory tracking and a swarm of WMRs can be described as the following equation.

$$e_i = \begin{bmatrix} \dot{\hat{r}}_i \\ \hat{r}_i \end{bmatrix} = \begin{bmatrix} \dot{r}_i - \dot{r}_{ri} \\ r_i - r_{ri} \end{bmatrix} \tag{4}$$

where:

$$r_{ri} = [ x_i \quad y_i \quad \theta \quad d_i ]^T$$

and the tracking error dynamics equation is given as:

$$\dot{e}_i = \begin{bmatrix} -M_i^{-1}(r_i)G_i(r_i, \dot{r}_i) & 0_{3 \times 3} \\ I_{3 \times 3} & 0_{3 \times 3} \end{bmatrix} e_i + \begin{bmatrix} -\ddot{r}_{ri} - M_i^{-1}(r_i)G_i(r_i, \dot{r}_i)\dot{r}_{ri} \\ 0_{3 \times 3} \end{bmatrix} + \begin{bmatrix} M_i^{-1}(r_i)O_i(r_i)\tau_i \\ 0_{3 \times 3} \end{bmatrix} \tag{5}$$

where:

$$\begin{bmatrix} \dot{e}_1 \\ \dot{e}_2 \\ \vdots \\ \dot{e}_n \end{bmatrix} = \begin{bmatrix} -M_1^{-1}(r_1)G_1(r_1, \dot{r}_1) & 0_{3 \times 3} & \dots & \dots & \dots & 0_{3 \times 3} \\ I_{3 \times 3} & 0_{3 \times 3} & \dots & \dots & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} & -M_2^{-1}(r_2)G_2(r_2, \dot{r}_2) & 0_{3 \times 3} & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} & I_{3 \times 3} & 0_{3 \times 3} & \dots & 0_{3 \times 3} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0_{3 \times 3} & \dots & \dots & \dots & -M_n^{-1}(r_n)G_n(r_n, \dot{r}_n) & 0_{3 \times 3} \\ 0_{3 \times 3} & \dots & \dots & \dots & I_{3 \times 3} & 0_{3 \times 3} \end{bmatrix} \begin{bmatrix} e_1 \\ e_2 \\ \vdots \\ e_n \end{bmatrix} + \begin{bmatrix} -\ddot{r}_{r1} - M_1^{-1}(r_1)G_1(r_1, \dot{r}_1)\dot{r}_{r1} \\ 0_{3 \times 3} \\ -\ddot{r}_{r2} - M_2^{-1}(r_2)G_2(r_2, \dot{r}_2)\dot{r}_{r2} \\ 0_{3 \times 3} \\ \vdots \\ -\ddot{r}_{rn} - M_n^{-1}(r_n)G_n(r_n, \dot{r}_n)\dot{r}_{rn} \\ 0_{3 \times 3} \end{bmatrix} + \begin{bmatrix} M_1^{-1}(r_1)O_1(r_1)\tau_1 \\ 0_{3 \times 3} \\ M_2^{-1}(r_2)O_2(r_2)\tau_2 \\ 0_{3 \times 3} \\ \vdots \\ M_n^{-1}(r_n)O_n(r_n)\tau_n \\ 0_{3 \times 3} \end{bmatrix}$$

To simplify the above equations, the tracking error dynamic equation can be described in the following format.

$$\dot{E} = F(r_1, r_2, \dots, r_n)E + C(r_1, r_2, \dots, r_n, \dot{r}_1, \dot{r}_2, \dots, \dot{r}_n) + \bar{U} \tag{6}$$

where the vector  $E$  is the tracking error between the desired trajectory tracking and a swarm of WMRs, both vector  $F$  and  $C$  are the dynamic model of a swarm of WMRs, and the vector  $\bar{U}$  is the nonlinear optimal control law.

$$E = [ e_1 \quad e_2 \quad \dots \quad e_n ]^T$$

$$\begin{aligned}
 & F(r_1, r_2, \dots, r_n) \\
 = & \begin{bmatrix} -M_1^{-1}(r_1)G_1(r_1, \dot{r}_1) & 0_{3 \times 3} & \dots & \dots & \dots & 0_{3 \times 3} \\ I_{3 \times 3} & 0_{3 \times 3} & \dots & \dots & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} & -M_2^{-1}(r_2)G_2(r_2, \dot{r}_2) & 0_{3 \times 3} & \dots & 0_{3 \times 3} \\ \dots & \dots & I_{3 \times 3} & 0_{3 \times 3} & \dots & 0_{3 \times 3} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0_{3 \times 3} & 0_{3 \times 3} & 0_{3 \times 3} & -M_n^{-1}(r_n)G_n(r_n, \dot{r}_n) & 0_{3 \times 3} & 0_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} & 0_{3 \times 3} & I_{3 \times 3} & 0_{3 \times 3} & 0_{3 \times 3} \end{bmatrix} \\
 C(r_1, r_2, \dots, r_n, r_{r1}, r_{r2}, \dots, r_{rn}) = & \begin{bmatrix} -\ddot{r}_{r1} - M_1^{-1}(r_1)G_1(r_1, \dot{r}_1)\dot{r}_{r1} \\ 0_{3 \times 3} \\ -\ddot{r}_{r2} - M_2^{-1}(r_2)G_2(r_2, \dot{r}_2)\dot{r}_{r2} \\ 0_{3 \times 3} \\ \vdots \\ -\ddot{r}_{rn} - M_n^{-1}(r_n)G_n(r_n, \dot{r}_n)\dot{r}_{rn} \\ 0_{3 \times 3} \end{bmatrix} \\
 \bar{U} = & \begin{bmatrix} M_1^{-1}(r_1)O_1(r_1)\tau_1 \\ 0_{3 \times 3} \\ M_2^{-1}(r_2)O_2(r_2)\tau_2 \\ 0_{3 \times 3} \\ \vdots \\ M_n^{-1}(r_n)O_n(r_n)\tau_n \\ 0_{3 \times 3} \end{bmatrix} \\
 = & \begin{bmatrix} M_1^{-1}(r_1)O_1(r_1) & 0_{3 \times 3} & \dots & \dots & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & \dots & \dots & \dots & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & M_2^{-1}(r_2)O_2(r_2) & \dots & \dots & \dots & 0_{3 \times 3} \\ 0_{3 \times 3} & \dots & \dots & \dots & \dots & 0_{3 \times 3} \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ 0_{3 \times 3} & \dots & \dots & \dots & \dots & M_n^{-1}(r_n)O_n(r_n)\tau_n \end{bmatrix} \begin{bmatrix} \tau_1 \\ \tau_2 \\ \vdots \\ \vdots \\ \vdots \\ \tau_n \end{bmatrix} = u_{i2}
 \end{aligned}$$

### 3. Problem Formulation and H<sub>2</sub> Controller Design

An analytic nonlinear H<sub>2</sub> control law for a swarm of WMRs will be inferred in this section. The design purpose of the trajectory tracking of a swarm of WMRs is to identify one nonlinear control law that fulfills the H<sub>2</sub> control performance index.

#### 3.1. Nonlinear H<sub>2</sub> Trajectory Tracking Problem

To give two weighted matrices, Q<sub>i2</sub> and R<sub>i2</sub>, the nonlinear H<sub>2</sub> trajectory tracking problem of a swarm of WMRs can be described with the H<sub>2</sub> control performance index, as below [28].

$$\begin{aligned}
 Y(u_{i2}^*) &= \min_{u_{i2}} Y(u_{i2}) = \min_{u_{i2}} \sum_{i=1}^n \left[ e_{i2}^T(t) Q_{i2} e_{i2}(t) + \int_0^{t_f} \left[ e_{i2}^T(t) Q_{i2} e_{i2}(t) + u_{i2}^T(t) R_{i2} u_{i2}(t) \right] dt \right] \\
 &= e_{i2}^T(0) J_{i2}(e_{i2}(0), 0) e_{i2}(0)
 \end{aligned} \tag{7}$$

Suppose the optimal robust control laws u<sub>i2</sub><sup>\*</sup> exist to satisfy Equation (7) for all t<sub>f</sub> ∈ [0, ∞] and the weighted matrices are Q<sub>i2</sub> = Q<sub>i2</sub><sup>T</sup> > 0 and R<sub>i2</sub> = R<sub>i2</sub><sup>T</sup>.

### 3.2. $H_2$ Control Design for a Swarm of WMRs

Base on the above tracking error dynamic Equation (5), the optimal  $H_2$  control law that analytically satisfies the  $H_2$  control performance index in Equation (7) can be expressed as:

$$u_{i2}^*(e, t) = -R_{i2}^{-1} \Delta_{i2}(e, t) J_{i2}(e, t) e_{i2}(t) \tag{8}$$

$$\Delta_{i2}(e, t) = \Gamma_{i2}^{-1} V M_{i2}^{-1}(r_i) \tag{9}$$

$$V = \begin{bmatrix} I_{3 \times 3} \\ 0_{3 \times 3} \end{bmatrix} \tag{10}$$

$$\Gamma_{i2} = \begin{bmatrix} \psi_{i2} I_{3 \times 3} & \phi_{i2} \\ 0_{3 \times 3} & I_{3 \times 3} \end{bmatrix} \tag{11}$$

where the positive scale  $\psi_{i2}$  and distinct positive matrix  $\phi_{i2} \in \mathfrak{R}^{3 \times 3}$  are constants.

If  $J_{i2}(e, t)$  in Equation (8) fulfills the as below nonlinear time-varying Riccati-like Equation (12), then this time-varying Riccati-like equation can solve the nonlinear  $H_2$  trajectory tracking problem of a swarm of WMRs in Equation (7).

$$\dot{J}_{i2}(e, t) + J_{i2}(e, t) W_{i2}(e, t) + W_{i2}^T(e, t) J_{i2}(e, t) + Q_{i2} - J_{i2}(e, t) \Delta_i(e, t) R_{i2}^{-1} \Delta_i^T(e, t) J_{i2}(e, t) = 0 \tag{12}$$

where

$$\begin{bmatrix} \dot{J}_{12}(e, t) + J_{12}(e, t) W_{12}(e, t) + W_{12}^T(e, t) J_{12}(e, t) + Q_{12} - J_{12}(e, t) \Delta_1(e, t) R_{12}^{-1} \Delta_1^T(e, t) J_{12}(e, t) \\ \dot{J}_{22}(e, t) + J_{22}(e, t) W_{22}(e, t) + W_{22}^T(e, t) J_{22}(e, t) + Q_2 - J_{22}(e, t) \Delta_2(e, t) R_2^{-1} \Delta_2^T(e, t) J_{22}(e, t) \\ \vdots \\ \vdots \\ \vdots \\ \dot{J}_{n2}(e, t) + J_{n2}(e, t) W_{n2}(e, t) + W_{n2}^T(e, t) J_{n2}(e, t) + Q_n - J_{n2}(e, t) \Delta_n(e, t) R_n^{-1} \Delta_n^T(e, t) J_{n2}(e, t) \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

and

$$W_{i2}(e, t) = \Gamma_{i2}^{-1} \begin{bmatrix} -M_{i2}^{-1}(r) G_{i2}(r, \dot{r}) & 0_{3 \times 3} \\ \frac{1}{\psi_{i2}} I_{3 \times 3} & -\frac{1}{\psi_{i2}} \phi \end{bmatrix} \Gamma_{i2}$$

Remark: Above the trajectory tracking problem of a swarm of WMRs for the analytic solution, one nonlinear time-varying Riccati-like equation can solve the  $H_2$  control performance index Equation (7) proofed in Appendix A.

From Equations (5) and (12), which are explicitly used to find out a closed-form solution,  $J_{i2}(e, t)$  of this nonlinear time-varying Riccati-like equations and nonlinear differential equations are certainly complicated tasks. In this research, the closed-form solution,  $J_{i2}(e, t)$ , will be directly derived from Equation (12), and this is an important contribution to the trajectory tracking design of a swarm of WMRs.

### 3.3. Closed-Form Solution of the Nonlinear Time-Varying Riccati-Like Equation

To consider the solution  $J_{i2}(e, t)$  of the nonlinear Riccati-like equation, Equation (12) can be shown in a more distinct form, with the state transformation in Equation (11) as:

$$J_{i2}(e, t) = \Gamma_{i2}^T \begin{bmatrix} M_{i2}(r_i) & 0_{3 \times 3} \\ 0_{3 \times 3} & Y_{i2} \end{bmatrix} \Gamma_{i2} \tag{13}$$

where  $\Gamma_{i2}$  and  $Y_{i2}$  are some designed positive distinct symmetric constant matrices and will be worked out from a pair of algebraic Riccati-like equations later.

Studying the second and third terms on the left-hand side of the time-varying Riccati-like Equation (12), and applying the tracking error dynamic system in Equation (5) and solution  $J_{i2}(e, t)$  in Equation (13), the following formulation can be described as:

$$J_{i2}(e, t)W_{i2}(e, t) + W_{i2}^T(e, t)J_{i2}(e, t) = \begin{bmatrix} 0_{3 \times 3} & Y_{i2} \\ Y_{i2} & 0_{3 \times 3} \end{bmatrix} + \Gamma_{i2}^T \begin{bmatrix} -\dot{M}_{i2}(r_i) & 0_{3 \times 3} \\ 0_{3 \times 3} & 0_{3 \times 3} \end{bmatrix} \Gamma_{i2} \quad (14)$$

The above equation can be verified easily as below:

$$V_{i2}^T(e, t)J_{i2}(e, t) = (\Gamma_{i2}^{-1}VM_{i2}^{-1}(r)) \Gamma_{i2}^T \begin{bmatrix} M_{i2}(r) & 0_{3 \times 3} \\ 0_{3 \times 3} & Y_{i2} \end{bmatrix} \Gamma_{i2} = V_{i2}^T \Gamma_{i2} \quad (15)$$

According to results in Equations (14) and (15), it is easy to confirm that Equation (12) can be represented as an algebraic Riccati-like equation:

$$\begin{bmatrix} 0_{3 \times 3} & Y_{i2} \\ Y_{i2} & 0_{3 \times 3} \end{bmatrix} + Q_{i2} - \Gamma_{i2}^T VR_{i2}^{-1}V^T \Gamma_{i2} = 0 \quad (16)$$

To assume one solution  $\Gamma_{i2}$  can be obtained mathematically and from Equation (15), the optimal robust control laws  $u_{i2}^*(e, t)$  in Equation (8) can be described as:

$$u_{i2}^*(e, t) = -R_{i2}^{-1}V^T \Gamma_{i2} e_{i2}(t) \quad (17)$$

By selecting

$$R_{i2} = \beta_{i2}^2 I_{6 \times 6} \quad (18)$$

where  $\beta_{i2} > 0$ .

To assume the weighted matrix  $Q_{i2}$  in Equation (7), it can be factorized by the Cholesky factorization, shown below:

$$Q_{i2} = \begin{bmatrix} Q_{112}^T Q_{112} & Q_{122} \\ Q_{212}^T & Q_{222}^T Q_{222} \end{bmatrix} \quad (19)$$

Applying the definitions of  $V$  and  $\Gamma_{i2}$  in Equations (10) and (11) together with assumption Equations (18) and (19), the Riccati-like Equation (16) can be separated into the following equations:

$$Q_{112}^T Q_{112} - \frac{1}{\beta_{i2}^2} \Gamma_{112}^T \Gamma_{112} = 0 \quad (20)$$

$$Y_{i2} + Q_{122} - \frac{1}{\beta_{i2}^2} \Gamma_{112}^T \Gamma_{122} = 0 \quad (21)$$

$$Y_{i2} + Q_{212}^T - \frac{1}{\beta_{i2}^2} \Gamma_{122}^T \Gamma_{112} = 0 \quad (22)$$

$$Q_{222}^T Q_{222} - \frac{1}{\beta_{i2}^2} \Gamma_{122}^T \Gamma_{122} = 0 \quad (23)$$

Based on Equations (20) and (23), the matrices  $\Gamma_{112}$  and  $\Gamma_{122}$  can be described as:

$$\Gamma_{112} = \beta_{i2} Q_{112} \quad (24)$$

and

$$\Gamma_{122} = \beta_{i2} Q_{222} \quad (25)$$

From  $\Gamma_{112}$  and  $\Gamma_{122}$  in Equations (24) and (25), the solution  $\Gamma_{i2}$  can be represented as:

$$\Gamma_{i2} = \begin{bmatrix} \beta_{i2}Q_{112} & \beta_{i2}Q_{222} \\ 0_{3 \times 3} & I_{3 \times 3} \end{bmatrix} \tag{26}$$

For fulfilling by  $\Gamma_{112} = \psi_{i2}I_{3 \times 3}$  in Equation (11), the weighted matrix  $Q_{112}$  in Equation (19) must be chosen as the below diagonal type:

$$Q_{112} = d_{112}I_{3 \times 3} \tag{27}$$

By selecting a positive scale,  $d_{112}$  and  $\psi_{i2}$ , they then have the relationship:

$$\psi_{i2} = \beta_{i2}d_{112} \tag{28}$$

By the same token, substituting Equations (26) into (21) and (22), we can obtain the following equation:

$$Q_{i2} = \begin{bmatrix} Q_{112}^T Q_{112} & Q_{122} \\ Q_{122}^T & Q_{222}^T Q_{222} \end{bmatrix} \tag{29}$$

Furthermore, the nonlinear  $H_2$  trajectory tracking problem of a swarm of WMRs is derived by the following  $H_2$  controller:

$$u_{i2}^* = -\frac{1}{\beta_{i2}} \begin{bmatrix} Q_{112} & Q_{222} \end{bmatrix} e_{i2}(t) \tag{30}$$

and the whole  $H_2$ -applied torques vector is described as below:

$$\tau_{i2} = S_{i2}(e, t) + \frac{1}{\psi_{i2}} u_{i2}^* \tag{31}$$

where

$$S_{i2}(e, t) = M_{i2}(r_i)(\ddot{r}_{ri} - \frac{Q_{222}}{d_{112}} \dot{r}) + G_{i2}(r, \dot{r})(\dot{r}_r - \frac{Q_{222}}{d_{112}} \dot{r}) \tag{32}$$

#### 4. Simulation Results and the Practical Implementation

Tracking a circular and S-type path with the proposed  $H_2$  method are presented by using the simulation Matlab software to verify the trajectory tracking performance of a swarm of WMRs.

##### 4.1. Set Up of Simulation Environments

We will apply the parameters of practical hardware of a swarm of WMRs as  $m = 5$  (kg),  $d = 7$  (cm),  $R = 8.9$  (cm), and  $r = 3.25$  (cm) to build up a simulation environment. This simulation scenario is close to the real situation, and one desired trajectory tracking circular and S-type for the simulation and verification is generated by the following equations.

Circular Trajectory:

$$\begin{cases} x = x_0 + r_d \cos(\theta_d) \\ y = y_0 + r_d \sin(\theta_d) \end{cases}$$

where  $r_d$  and  $\theta_d = \int_0^t \omega_d dt$  are the radius of the desired circular trajectory and desired rotation angle with a constant predefined angular velocity  $\omega_d$ , respectively. The initial conditions of the desired trajectory are  $x_0 = 0$ (m),  $y_0 = 0$ (m), and  $\omega_d = 3^\circ/s$ . The control of a swarm of WMRs, which is driven by the proposed method, starts from initial conditions, as below.

Circular Scenario:

$$x_c(0) = 1.9(\text{m}), y_c(0) = 0(\text{m}), \theta(0) = 0^\circ; x_c(0) = -1.9(\text{m}), y_c(0) = 0(\text{m}), \text{ and } \theta(0) = 0^\circ$$

$$x_c(0) = 0(\text{m}), y_c(0) = 1.9(\text{m}), \theta(0) = 0^\circ; x_c(0) = 0(\text{m}), y_c(0) = -1.9(\text{m}), \text{ and } \theta(0) = 0^\circ$$



S type Trajectory:

$$\begin{cases} x = x_0 + r_d \cos(2\theta_d) \\ y = y_0 + 2r_d \sin(\theta_d) \end{cases}$$

where  $r_d$  and  $\theta_d = \int_0^t \omega_d dt$  are the radius of the desired S-type trajectory and desired rotation angle with a constant predefined angular velocity  $\omega_d$ , respectively. As a circular scenario, the initial conditions of the desired trajectory are  $x_0 = 0(m)$ ,  $y_0 = 0(m)$ , and  $\omega_d = 3^\circ /s$ . This case will also verify the trajectory tracking ability of this proposed  $H_2$  control design. The control of a swarm of WMRs, which is driven by the proposed method, starts with the following initial locations.

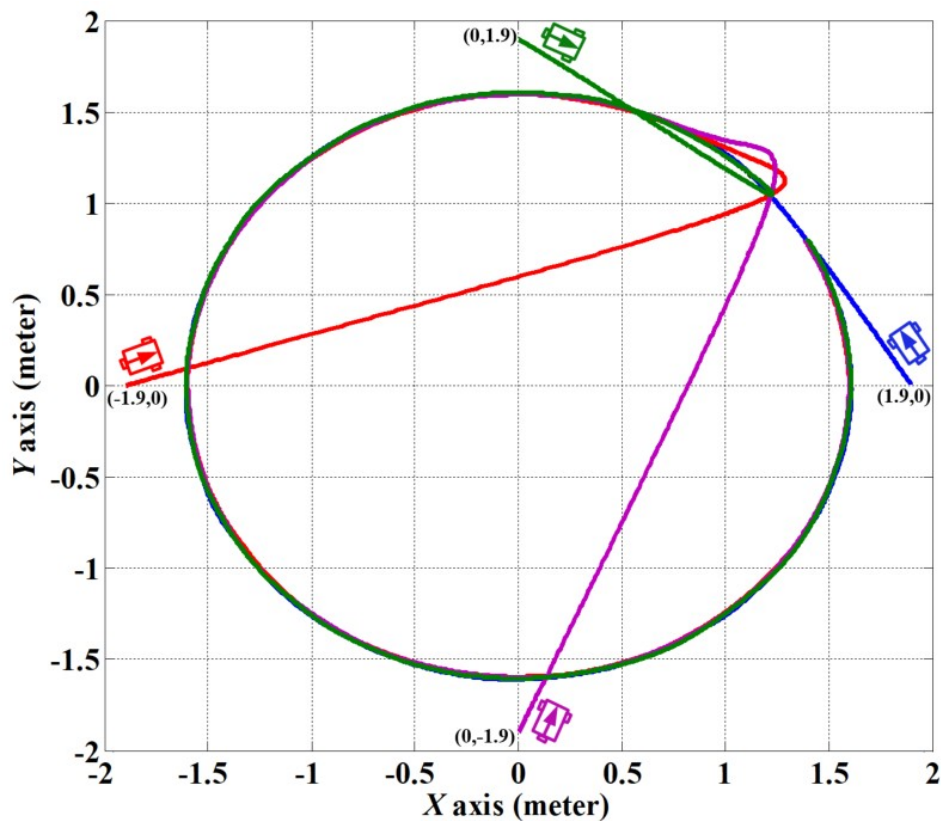
S type Scenario:

$$x_c(0) = 1.9(m), y_c(0) = 0(m), \theta(0) = 0^\circ; x_c(0) = -1.9(m), y_c(0) = 0(m), \text{ and } \theta(0) = 0^\circ$$

$$x_c(0) = 0(m), y_c(0) = 1.9(m), \theta(0) = 0^\circ; x_c(0) = 0(m), y_c(0) = -1.9(m), \text{ and } \theta(0) = 0^\circ$$

#### 4.2. Simulation Results

Figures 2–10 display the simulation results of the swarm of WMRs driven by  $H_2$  control design for tracking a desired circular-type trajectory with a 1.6 m radius. Figure 2 shows the tracking result of the  $H_2$  control design in the x-y axis concerning the predefined circular trajectory. As the depicted result, the trajectory tracking the performance of the  $H_2$  control design to the desired circular trajectory is an achievement. From Figures 3–6, it is significant with tracking errors in the x-y axis and angle. From Figures 7–10, the trajectory tracking torques are significant convergence rates, which approach zero quickly. This proposed  $H_2$  control design can drive a swarm of WMRs to track the desired circular trajectory quickly and achieve better trajectory tracking performance no matter the different positions or the rotation angles.



**Figure 2.** Circular trajectory tracking result of  $H_2$  proposed control design with respect to four different initial conditions: ( $x_c = 1.9$  m,  $y_c = 0$  m, blue line), ( $x_c = -1.9$  m,  $y_c = 0$  m, red line), ( $x_c = 0$  m,  $y_c = 1.9$  m, green line), and ( $x_c = 0$  m,  $y_c = -1.9$  m, purple line).

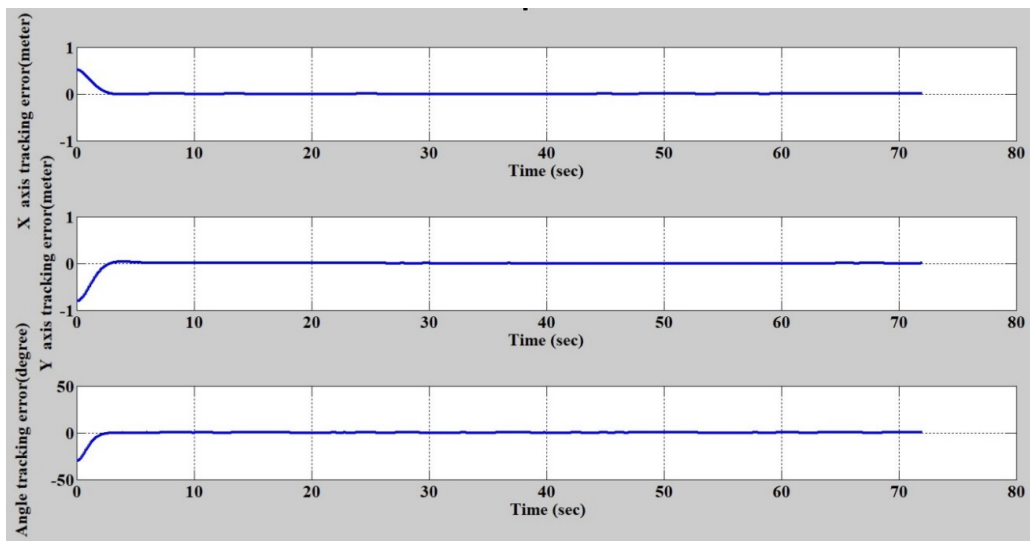


Figure 3. Results of circular trajectory tracking errors concerning initial point ( $x_c = 1.9$  m,  $y_c = 0$  m).

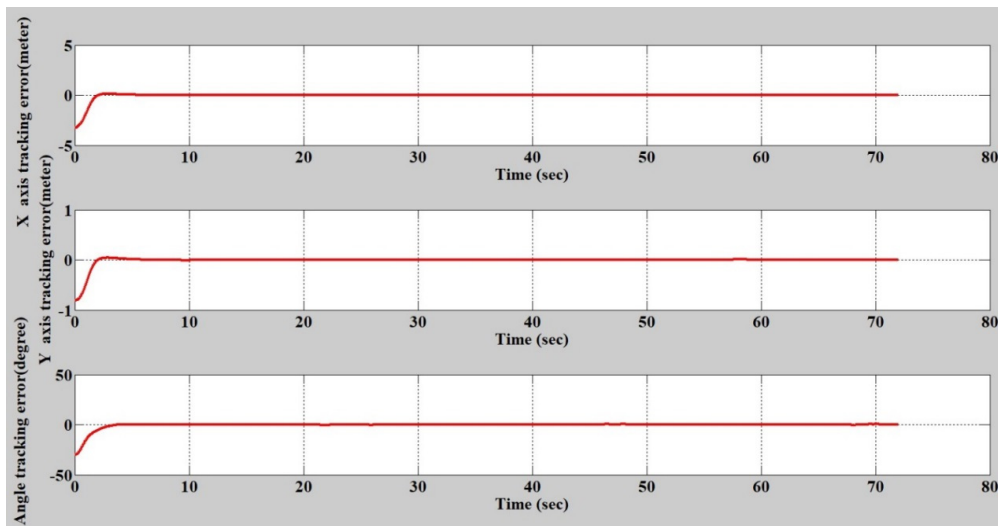


Figure 4. Results of circular trajectory tracking errors concerning initial point ( $x_c = -1.9$  m,  $y_c = 0$  m).

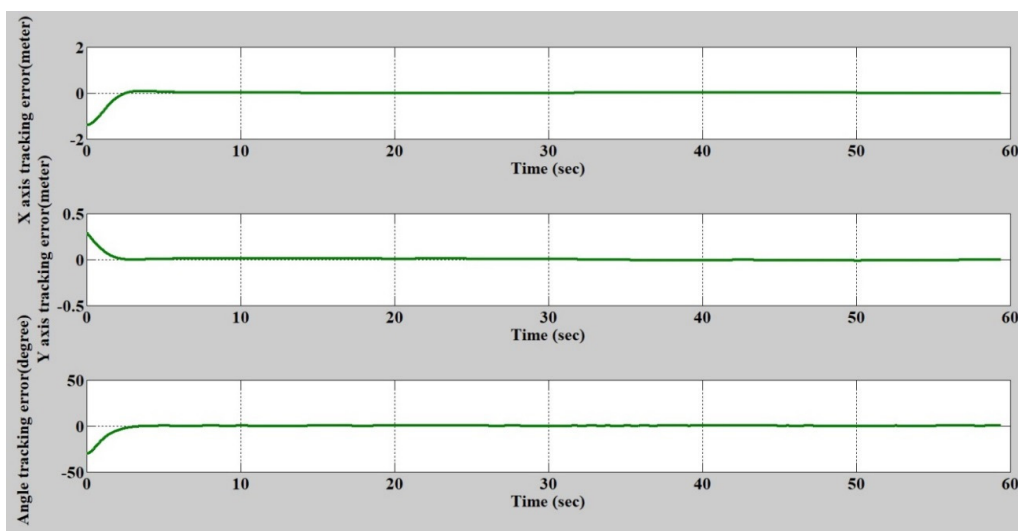


Figure 5. Results of circular trajectory tracking errors concerning initial point ( $x_c = 0$  m,  $y_c = 1.9$  m).

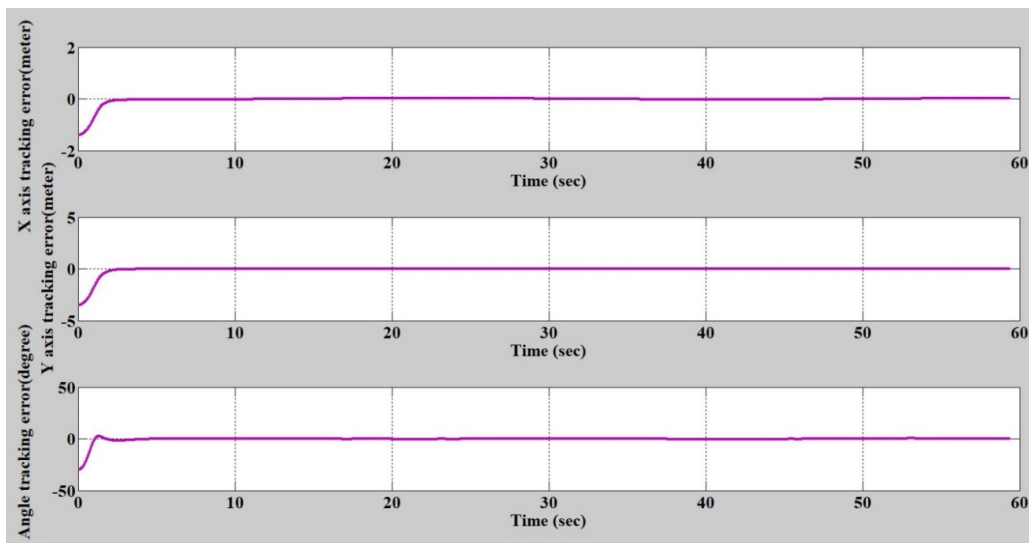


Figure 6. Results of circular trajectory tracking errors concerning initial point ( $x_c = 0$  m,  $y_c = -1.9$  m).

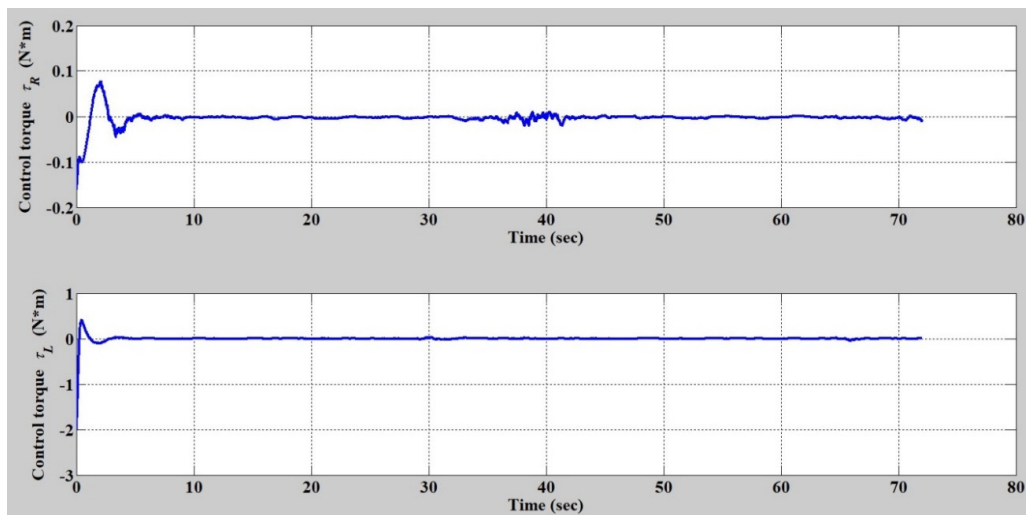


Figure 7. Results of circular trajectory tracking torques concerning initial point ( $x_c = 1.9$  m,  $y_c = 0$  m).

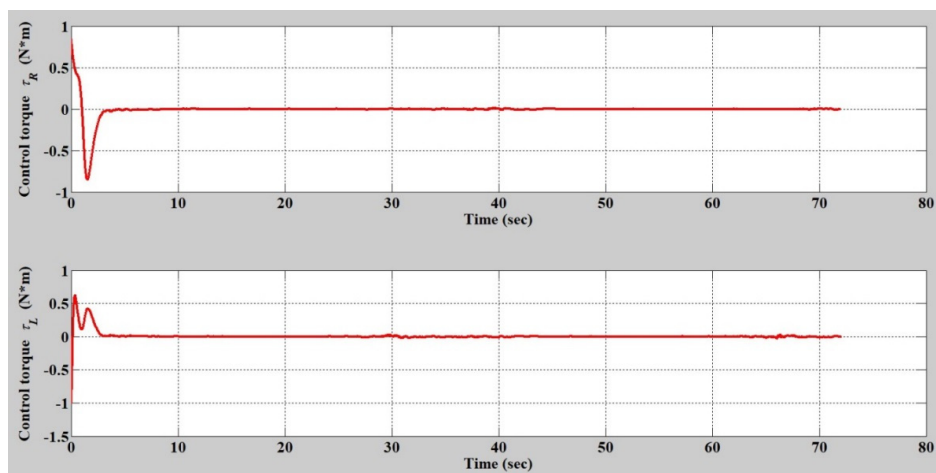


Figure 8. Results of circular trajectory tracking torques concerning initial point ( $x_c = -1.9$  m,  $y_c = 0$  m).

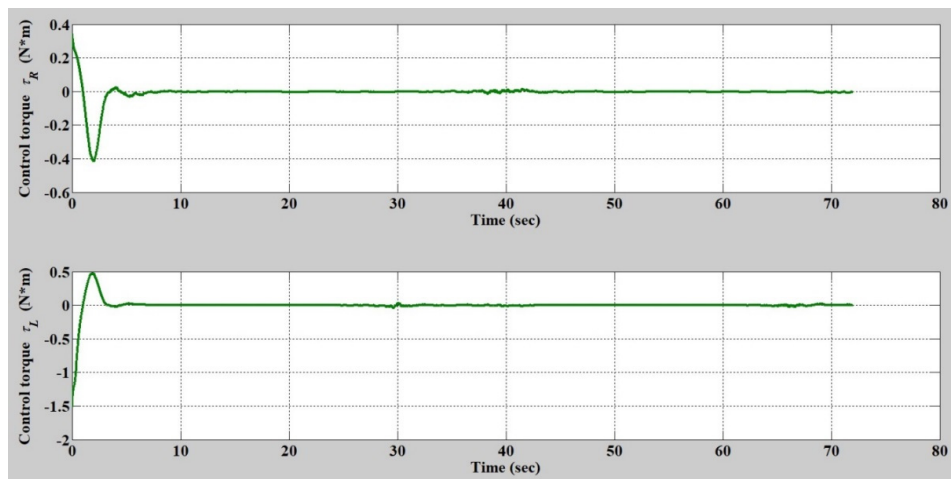


Figure 9. Results of circular trajectory tracking torques concerning initial point ( $x_c = 0$  m,  $y_c = 1.9$  m).

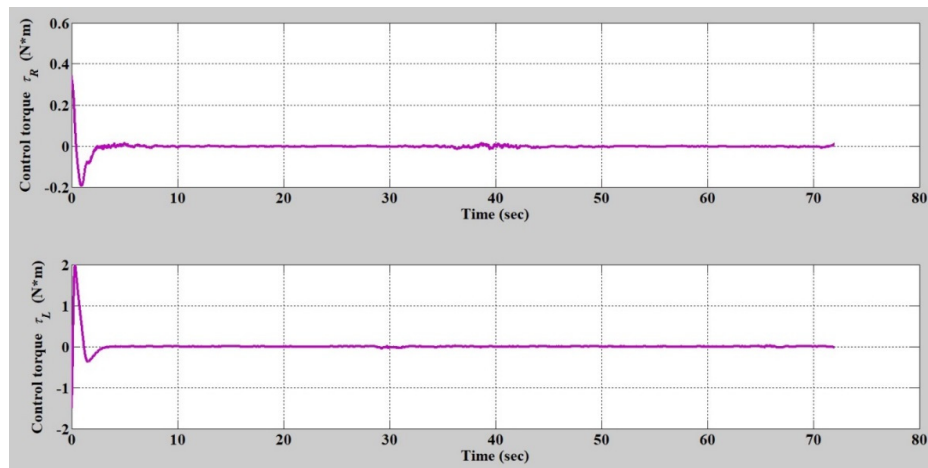
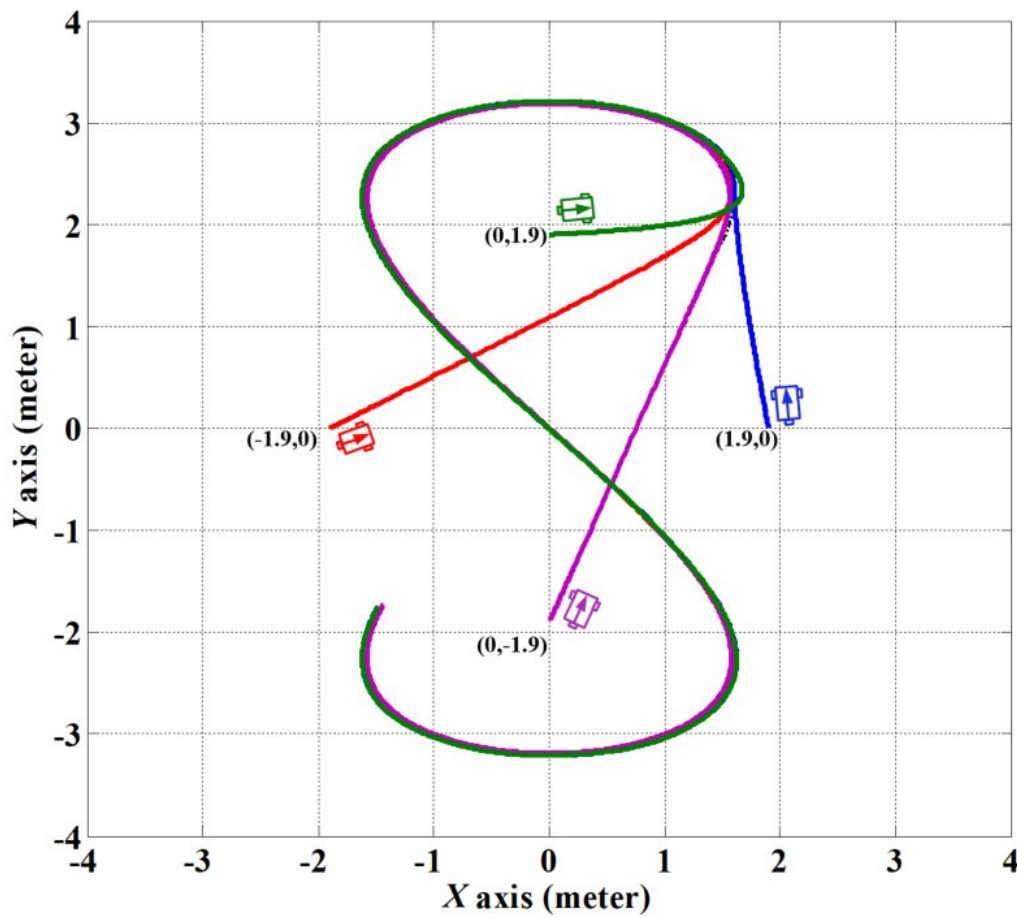
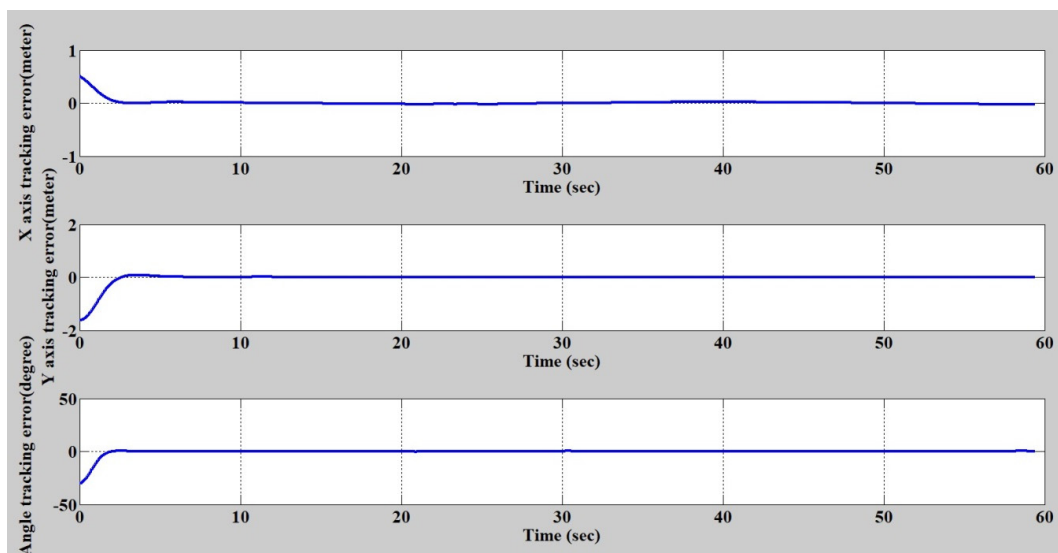


Figure 10. Results of circular trajectory tracking torques concerning initial point ( $x_c = 0$  m,  $y_c = -1.9$  m).

Finally, the  $H_2$  control design of a desired S-type trajectory with a 1.6 m radius gives the simulation results of Figures 11–19. Figure 11 reveals the trajectory tracking performance of the  $H_2$  control design in the desired S-type trajectory. From the significant effect, the trajectory tracking performance of this proposed  $H_2$  design to the reference S-type trajectory also demonstrates an excellent outcome. From Figures 12–19, it points out noticeable results, and this intended  $H_2$  design can track the desired S-type trajectory fast, and errors and torques convergence rate approach zero quickly.



**Figure 11.** S type trajectory tracking result of  $H_2$ -proposed control design with respect to four different initial conditions:  $(x_c = 1.9 \text{ m}, y_c = 0 \text{ m})$ ,  $(x_c = -1.9 \text{ m}, y_c = 0 \text{ m})$ ,  $(x_c = 0 \text{ m}, y_c = 1.9 \text{ m})$ , and  $(x_c = 0 \text{ m}, y_c = -1.9 \text{ m})$ , purple line).



**Figure 12.** Results of S-type trajectory tracking errors concerning initial point  $(x_c = 1.9 \text{ m}, y_c = 0 \text{ m})$ .

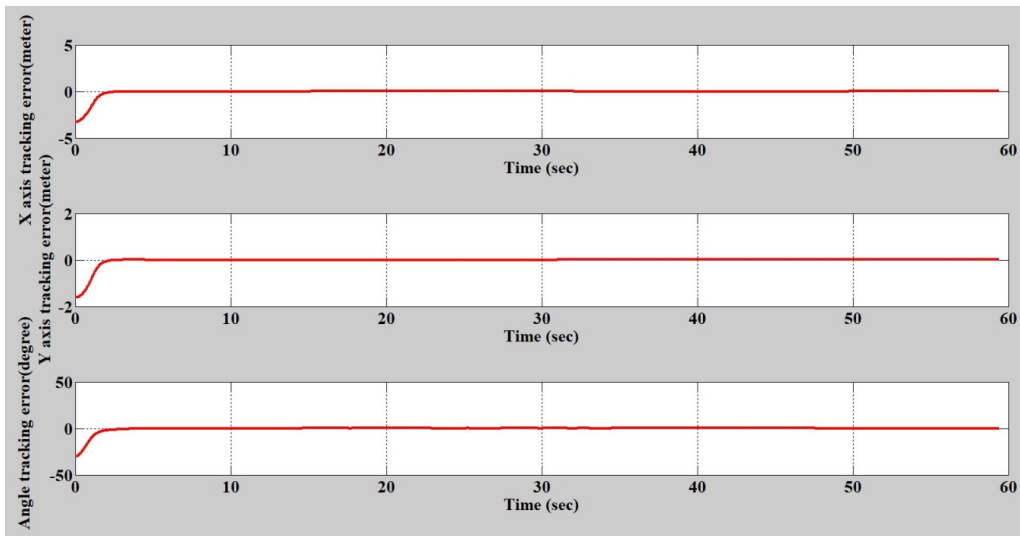


Figure 13. Results of S-type trajectory tracking errors concerning initial point ( $x_c = -1.9$  m,  $y_c = 0$  m).

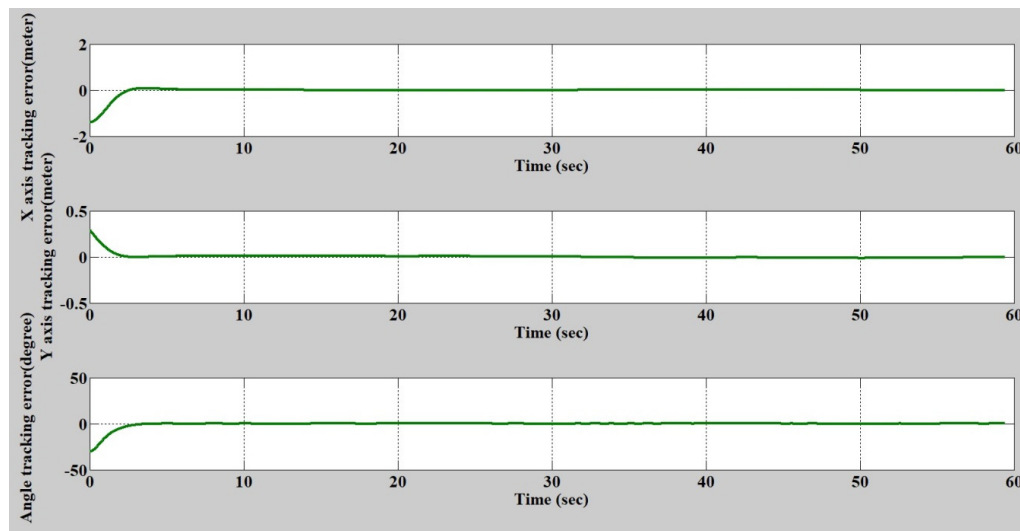


Figure 14. Results of S-type trajectory tracking errors concerning initial point ( $x_c = 0$  m,  $y_c = 1.9$  m).

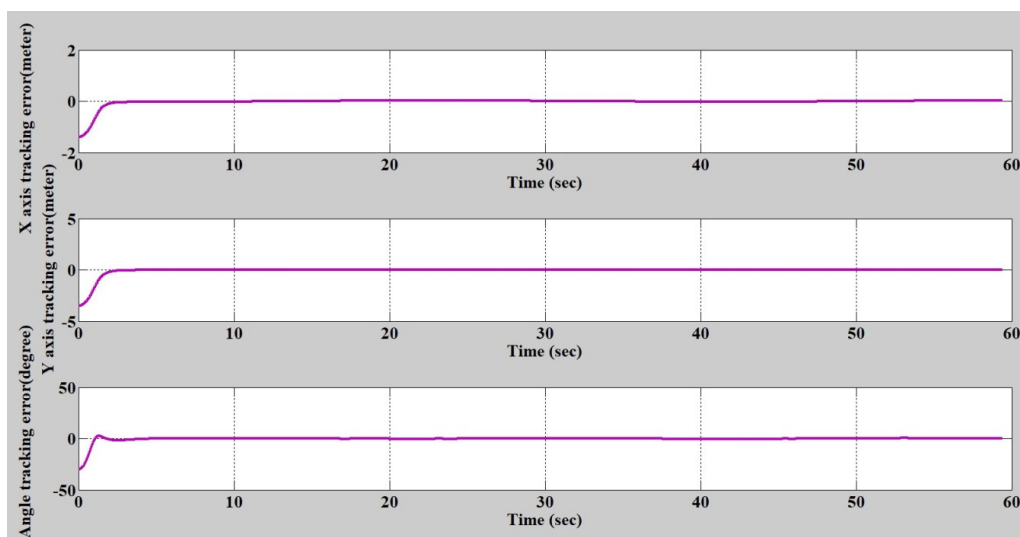


Figure 15. Results of S-type trajectory tracking errors concerning initial point ( $x_c = 0$  m,  $y_c = -1.9$  m).

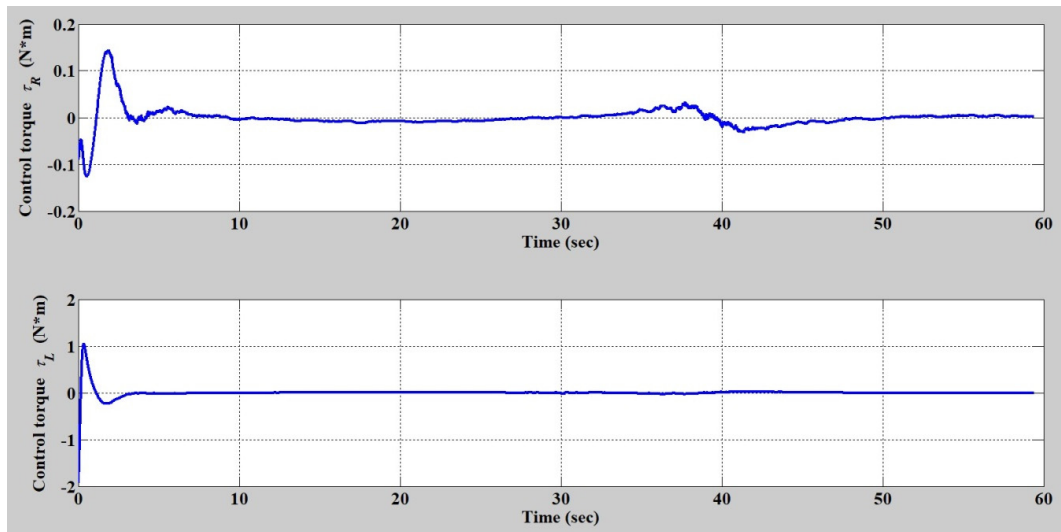


Figure 16. Results of S-type trajectory tracking torques concerning initial point ( $x_c = 1.9$  m,  $y_c = 0$  m).

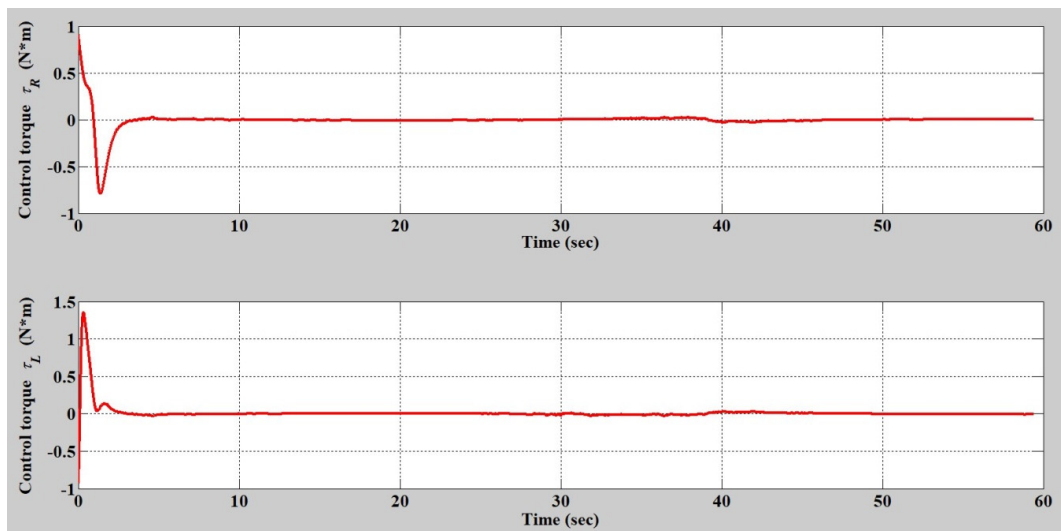


Figure 17. Results of S-type trajectory tracking torques concerning initial point ( $x_c = -1.9$  m,  $y_c = 0$  m).

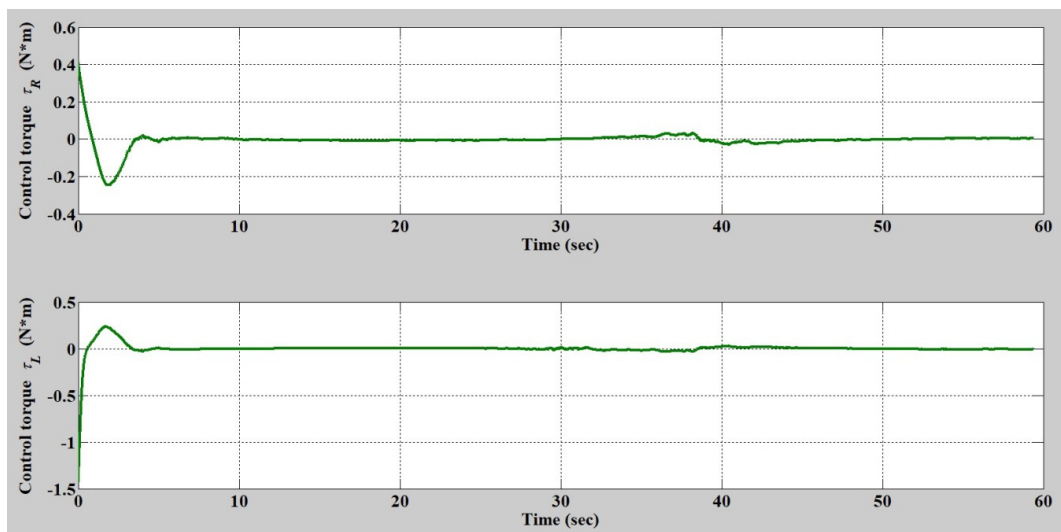
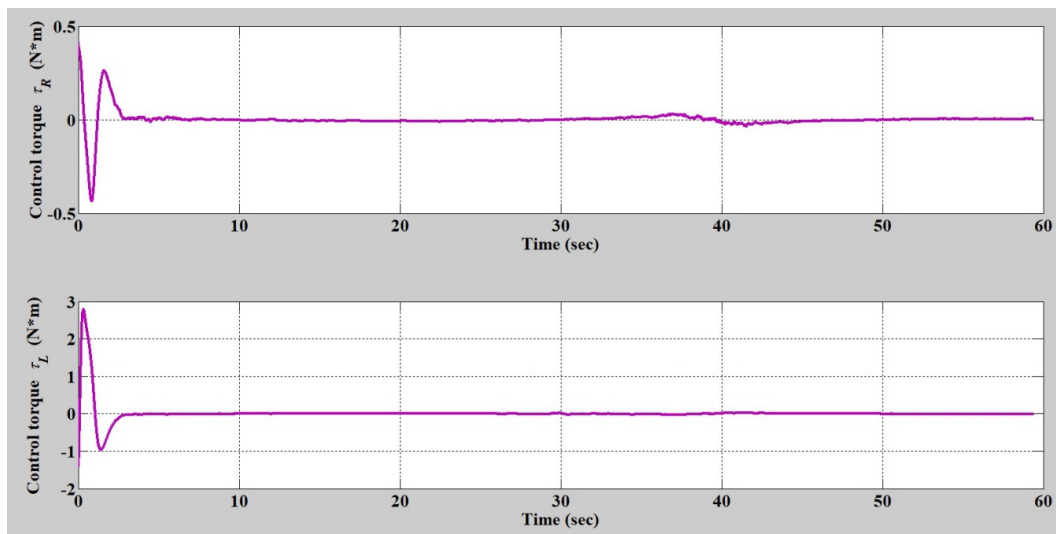


Figure 18. Results of S-type trajectory tracking torques concerning initial point ( $x_c = 0$  m,  $y_c = 1.9$  m).



**Figure 19.** Results of S-type trajectory tracking torques concerning initial point ( $x_c = 0$  m,  $y_c = -1.9$  m).

## 5. Conclusions

Based on existing published documentation and regular literature, lots of research results with passable optimal performances, and very complicated control methodologies to neural networks, feedback linearization, sliding mode control, and backstepping methodologies for the trajectory tracking subject of a swarm of WMRs have been investigated. For enhancing these improvable control properties, a nonlinear  $H_2$  control method that is analytically inferred is a global solution without any approximations. In this paper, the nonlinear trajectory tracking subject of a swarm of WMRs is completely constructed to promote the tracking capability and reduce the cost of the effort of design, implementation, and computational consumption. Especially, it is actually a simple and easy-to-implement control method. Incidentally, it is challenging to find out this control method due to a solution of one nonlinear time-varying differential equation that needs to satisfy the  $H_2$  performance index. Fortunately, an  $H_2$  closed-form solution of this nonlinear trajectory tracking subject can be inferred directly by applying to an appropriate mathematical conversion method. According to the results of the simulation and practical experiment, this nonlinear  $H_2$  control method not only fulfills the simulation results for tracking the predefined trajectories in a circular and S-type scenario, but also achieves perfect and engaged trajectory tracking performance in the real world. Finally, we can also assert that this nonlinear  $H_2$  control method gives some outstanding benefits when a swarm of WMRs performs trajectory tracking missions, such as security, transportation, patrol, and so forth.

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## Appendix A

Proof of  $H_2$  performance index with nonlinear time-varying Riccati-like equation.



**Proof.** First, consider the cost function  $Y_2(u_{i2})$ . It was evident that Equation (7) can be rewritten as:

$$Y_2(u_{i2}) = \min_{u_{i2}} \sum_{i=1}^n \left[ e_{i2}^T(0) J_{i2}(e_{i2}(0), 0) e_{i2}(0) + \int_0^{t_f} \left[ e_i^T(t) \left( \dot{J}_{i2}(e, t) + J_{i2}(e, t) W_{i2}(e, t) + W_{i2}^T(e, t) J_{i2}(e, t) + Q_{i2} \right) e_{i2}(t) + u_{i2}^T(t) R_{i2} u_{i2}(t) + u_{i2}^T(t) \Delta_{i2}^T(e, t) J_{i2}(e, t) e_{i2}(t) + e_{i2}^T(t) J_{i2}(e, t) \Delta_{i2}(e, t) u_{i2}(t) \right] dt \right] \quad (A1)$$

By some manipulations and using Riccati-like Equation (12), we obtain:

$$Y_2(u_{i2}, 0) = \min_{u_{i2}} \sum_{i=1}^n \left[ e_{i2}^T(0) J_{i2}(e_{i2}(0), 0) e_{i2}(0) + \int_0^{t_f} \left[ \left( u_{i2}(t) + R_{i2}^{-1} \Delta_{i2}^T(e, t) J_{i2}(e, t) e_{i2}(t) \right)^T R_{i2} \left( u_{i2}(t) + R_{i2}^{-1} \Delta_{i2}^T(e, t) J_{i2}(e, t) e_{i2}(t) \right) \right] dt \right] \quad (A2)$$

Then, by choosing control law like that in Equation (8), we can conclude that:

$$Y_2(u_{i2}^*(t)) = \min_{u_{i2}} \sum_{i=1}^n J_{i2}(u_{i2}) = \min_{u_{i2}} \sum_{i=1}^n \left[ e_{i2}^T(0) J_{i2}(e_{i2}(0), 0) e_{i2}(0) \right] \quad (A3)$$

This is the  $H_2$  control performance index in Equation (7).  $\square$

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