

Article

# Fully Implantable Neural Stimulator with Variable Parameters

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**Abstract:** Neural implantable systems have promoted the development of neurosurgery research and clinical practice. However, traditional tethered neural implants use physical wires for power supply and signal transmission, which have many restrictions on implant targets. Therefore, untethered, wireless, and controllable neural stimulation has always been widely recognized as the engineering goal of neural implants. In this paper, magnetically coupled resonant wireless power transfer (MCR-WPT) technology is adopted to design and manufacture a wireless stimulator for the electrical stimulation experiment of nerve repair. In the process of device development, SCM technology, signal modulation, demodulation, wireless power supply, and integration/packaging are used. Through experimental tests, the stimulator can output single-phase pulse signals with a variable frequency of (1–20 Hz), a duty cycle of (1–50%), and voltage. The average power is approximately 25 mW. The minimum pulse width of the signal is 200  $\mu$ s and the effective distance of transmission is 1–4 cm. The stimulator can perform low-frequency, safe and controllable wireless stimulation.

**Keywords:** electrical stimulation; magnetic coupling resonance; implantable neural stimulator; wireless control



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## 1. Introduction

Peripheral nerve injury is a major problem in public health, accounting for 2–5% of all trauma cases [1]. Peripheral nerve injury can lead to sensory, nutritional, and motor disturbances in the innervated area.

At present, the common repair methods for nerve injury include cell therapy [2], molecular therapy [3], and catheter therapy [4]. Additionally, studies have proved that electrical stimulation therapy can accelerate nerve regeneration after injury and promote the recovery of corresponding motor functions [5], which is characterized by good effects, small side effects, and clear action sites. Electrical stimulation can be widely used in clinical and neurological therapy and rehabilitation, and experimental electrical stimulation in animal models can be used for the preclinical evaluation of the safety and effectiveness of implantable neural prostheses, as well as for basic research on neural function and behavior. The stimulator designed in this study is modeled on rats and other small animals.

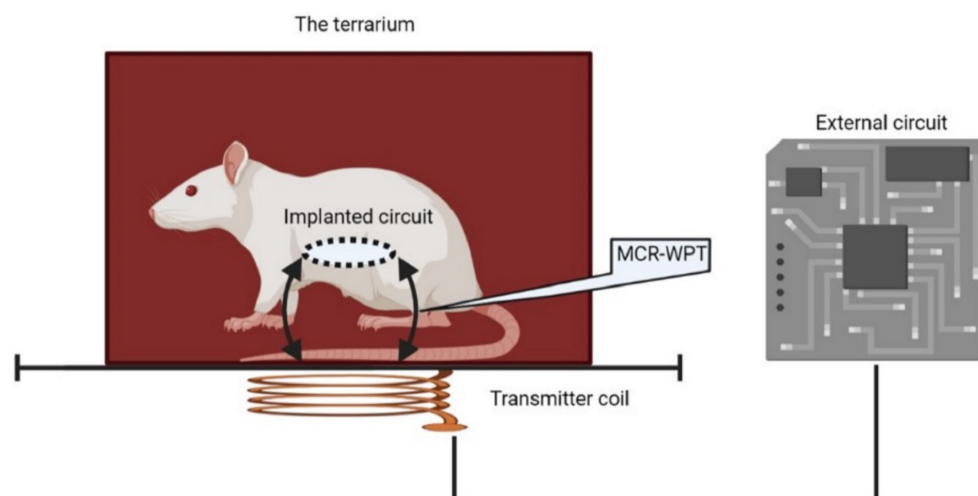
In animal model experimental studies, methods of wired percutaneous connections between arrays and external stimulators is common and convenient [6–10]. However, those methods have many disadvantages: (1) they restrict movement and natural behavior; (2) the wound has a high risk of potential infection; (3) the mechanical failure due to mechanical stress of exposed equipment and impact of animals. Current studies have proposed remote control neural stimulators for the movement control of small animals [11–14]. These devices can eliminate binding lines and enable animals to move unrestricted. However, there is still a percutaneous connection between the implanted cortical array and the stimulator. Battery-powered implantable stimulators can solve the problem [15], although their power capacity is limited and the use of fully implantable wireless stimulators have been demonstrated through research. The electronic components are inserted into the subcutaneous space around the scapula or head without any components outside the body [16,17]. The implant

is small, light in weight, and long in life, but the frequency and intensity of the pulse stimulation are not effectively controlled.

Fortunately, the past decade has witnessed significant technical advancement on how to achieve wireless neural implants. The research of electric stimulation in animal experiments has made breakthroughs in the fields of motor control, neural decoding, and neural therapy. For example, the movement of some living insects can be controlled by implantable electrical nerve stimulation, and the speed can be adjusted by varying the frequency of stimulation [18–21]. The robotic insects could be used for search and rescue in complex environments. Moreover, Ho, J. S. et al. [22] demonstrated a wirelessly powered method for powering tiny implants almost anywhere in the body. Kate et al. [23] developed a miniature wireless stimulator to achieve photo genetic manipulation of neural circuits in the nervous system. A tiny LED was implanted into mice, and optical stimulation with blue light in the spinal cord and cerebral cortex elicited circling behavior and increased the speed of locomotion. Thanks to technological advances and the success of clinical trials, fully implantable solutions have become a reality and are now available for the treatment of nerve injury [24]. These studies provide references for our design.

Based on the above discussion, the design criteria can be proposed as follows: (1) meet the basic requirements of bioelectric stimulation; (2) precise and adjustable stimulus parameters; (3) wireless control and drive; (4) miniaturization; (5) long-term power supply capability.

Given these objectives, a novel wireless, fully implantable small neural electrical stimulator has been designed to deliver long-term, variable parameter pulse stimulation to experimental targets under natural conditions (Figure 1). This work can overcome the above technical shortcomings.



**Figure 1.** A fully implantable electrical nerve stimulator: the implanted circuit with a rectifier and receive coil ( $T_X$ ), the transmit coil ( $R_X$ ) under the terrarium supplies energy by using magnetically coupled resonant wireless power transfer (MCR-WPT), the external circuit is used to generate signals.

Table 1 shows the comparisons between existing research results with our work in stimulus parameters. Lu et al. [25] and Chen et al. [26] show that the stimulation current at low frequency and milliampere-level have a significant effect on promoting nerve fiber regeneration. The optimal frequency they used was below 20 Hz, and our stimulator is designed to achieve a frequency regulation of 1–20 Hz. As for single-phase stimulation and dual-phase stimulation, Chen et al. [26] show that the difference in waveform characteristics can be ignored.

The stimulator transmission distance is set to a few centimeters and the coupling frequency is set to 3 MHz. At this frequency, the energy loss associated with bioliquid absorption can be reduced [28].



In the following chapters, we introduce the circuit model manufacturing method and conduct simulation analyses and experimental verifications.

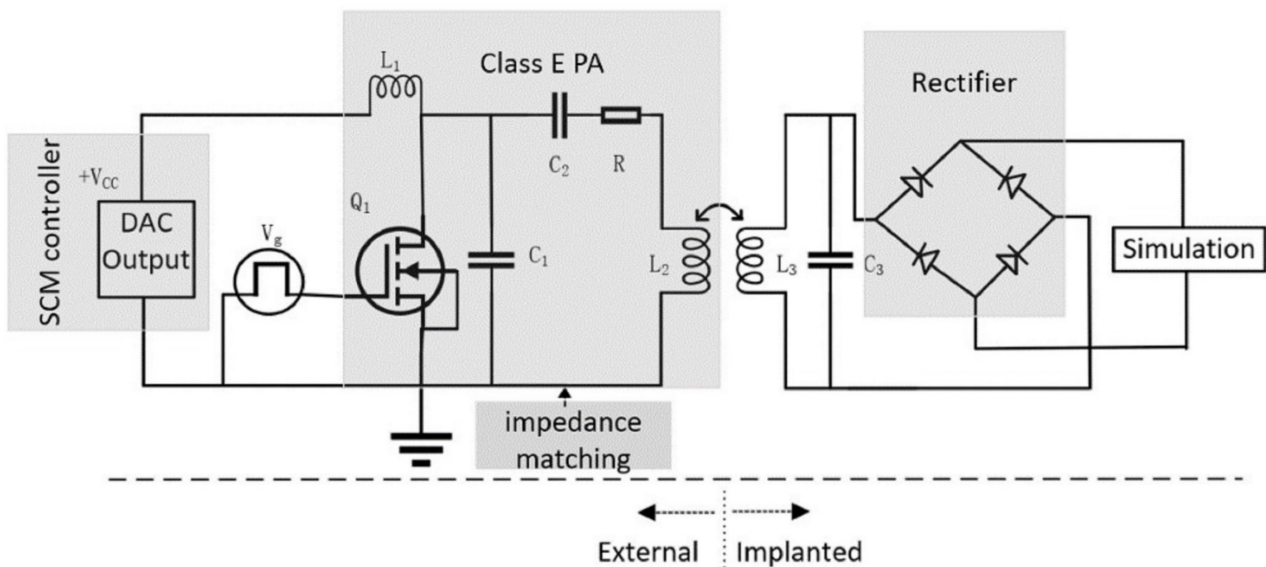
**Table 1.** The comparisons between existing research results with our work in stimulus parameters.

	[7]	[12]	[17]	[27]	This Work
Waveform	Irregular	Single-phase pulse	Dual-phase pulse	Irregular	Single-phase pulse
Frequency	0, 10, 20, 80 Hz	100 Hz	60 Hz	1–255 Hz (1-Hz step)	1–20 Hz (1-Hz step)
Duty cycle	N/A	50%	1.2%	N/A	1–50%
Output voltage	11 V	9 V	N/A	0.1–10 V	0–5 V
Current	0–5 mA	15 mA	0.6 mA	1 mA	5 mA
Average power	30 mW	67.5 mW	N/A	10 mW	25 mW

## 2. Materials and Methods

### 2.1. System Overview

The system consists of an implanted stimulator and an external circuit which are wirelessly connected by MCR-WPT (Figure 2). The external circuit includes a SCM (Single-Chip Micryoco) controller, impedance matching network, and the Class E PA (Power Amplifier). The assembly of the implanted stimulator consists of a rectifier and the stimulating circuit. The  $L_2$  and  $L_3$  are the transmit (Tx) and receive (Rx) coil in the MCR-WPT system, respectively. The pulse  $V_g$  is the gate drive signal. The impedance matching network will be covered in the next section.



**Figure 2.** The system overview of the wireless stimulation system.

### 2.2. Static Parameter Analysis

#### 2.2.1. The External Circuit

In this work, the Class E PA is used as a power source. Among them, ZVS (Zero Voltage Switching) Class E PA has the highest working efficiency [29]. When the transistor is on, the capacitor  $C_1$  is short-circuited, and  $R, C_2, L_2$  form a series resonant circuit. The resonance frequency  $f_1$  and q-factor  $Q_1$  can be described as:

$$f_1 = \frac{1}{2\pi\sqrt{L_2C_2}} \tag{1}$$

$$Q_1 = \frac{\omega_1 L_2}{R} = \frac{1}{\omega_1 C_2 R} \tag{2}$$

When the transistor is disconnected,  $C_1$ ,  $R$ ,  $C_2$ ,  $L_2$  form a series resonant circuit. The resonance frequency  $f_2$  and q-factor  $Q_2$  can be described as:

$$f_2 = \frac{1}{2\pi\sqrt{L_2\frac{C_1C_2}{C_1+C_2}}} \quad (3)$$

$$Q_2 = \frac{\omega_2 L_2}{R} = \frac{1}{\omega_2 R C_1 C_2 / (C_1 + C_2)} \quad (4)$$

To ensure the efficiency of class E PA, the signal frequency ( $f_e$ ) of the driving transistor switch should be between  $f_1$  and  $f_2$ . We assume that the duty cycle is  $D = 0.5$  and the choke inductance  $L_1$  is large enough. When the supply voltage  $V_{CC}$ , the load resistance  $R_0$ , the output power delivered to the load  $P_O$ , the operating frequency  $f_e$  and the q-factor  $Q_L$  is given, we can calculate circuit component parameter values from the following formula [30].

$$R = 0.5768 \frac{V_{CC}^2}{P_O} \left( 1.001245 - \frac{0.452}{Q} - \frac{0.4}{Q^2} \right) \quad (5)$$

$$L_2 = \frac{QR}{\omega} \quad (6)$$

$$C_1 = \frac{1}{\omega R} \left[ \frac{8}{\pi(\pi^2 + 4)} \left( 0.999 + \frac{0.914}{Q} - \frac{1.03}{Q^2} \right) + \frac{0.6}{Q(L_1/L_2)} \right] \quad (7)$$

$$C_2 = \frac{1}{\omega R} \left[ \left( \frac{1}{Q - 0.105} \right) \left( 1.001 + \frac{1.015}{Q - 1.788} \right) - \frac{0.2}{Q(L_1/L_2)} \right] \quad (8)$$

Referring to circuit specifications described in [29,30], we set  $R_0 = 50 \Omega$ ,  $P_O = 1 \text{ W}$ ,  $f_e = 3 \text{ MHz}$ , and  $Q_L = 10$  here. As the supply voltage of the stimulator is adjustable, and  $R$  varies from 0 to  $15 \Omega$ , we set  $R = 10 \Omega$ . As this value is different from the actual load resistance  $R_0$ , we can use the  $\pi$  type impedance matching network to transform  $R_0$  to  $R$ . According to the design indicators in Table 1, the final component parameters are shown in Table 2. There is a comparison,  $f_1 = 2.81 \text{ MHz} < f_e = 3 \text{ MHz} < f_2 = 3.48 \text{ MHz}$ , which conforms to the design index.

**Table 2.** The value of the circuit component.

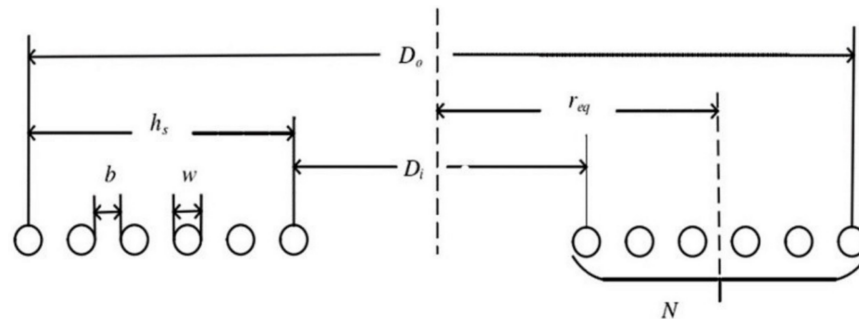
Items	Values	Items	Values
$C_1$	1.62 nF	$L_1$	200 $\mu\text{H}$
$C_2$	0.87 nF	$L_2$	3.64 $\mu\text{H}$
$C_3$	494 pF	$L_3$	5.59 $\mu\text{H}$
$C_4$	1.63 nF	$L_4$	0.88 $\mu\text{H}$
$C_5$	1.37 nF	$R$	10 $\Omega$

### 2.2.2. The Implanted Stimulator

The key point of the implanted stimulator is the design of the  $R_x$  coil  $L_3$ . In our work, it adopts the manufacturing model of a planar spiral coil, and its cross-section is shown in Figure 3. The value of  $L$  can be calculated using the following formula [31]:

$$L = \frac{N^2(D_0 - N(w + b))^2}{16D_0 + 28N(w + b)} \times \frac{39.37}{10^6} (H) \quad (9)$$

where  $D_0$  is the outer diameter of the coil;  $N$  is the number of turns of the coil;  $w$  and  $b$  denote the diameter of wire and the spacing between the wires, respectively.



**Figure 3.** The cross-section of the planar spiral coil. The  $D_i$  is the inner diameter; the  $r_{eq}$  is the equivalent radius of the coil; the  $h_s$  is the width of the coil.

Moreover,  $L_3$  and  $C_3$  resonate at the operating frequency  $f_e = 3$  MHz and their values must satisfy the formula:

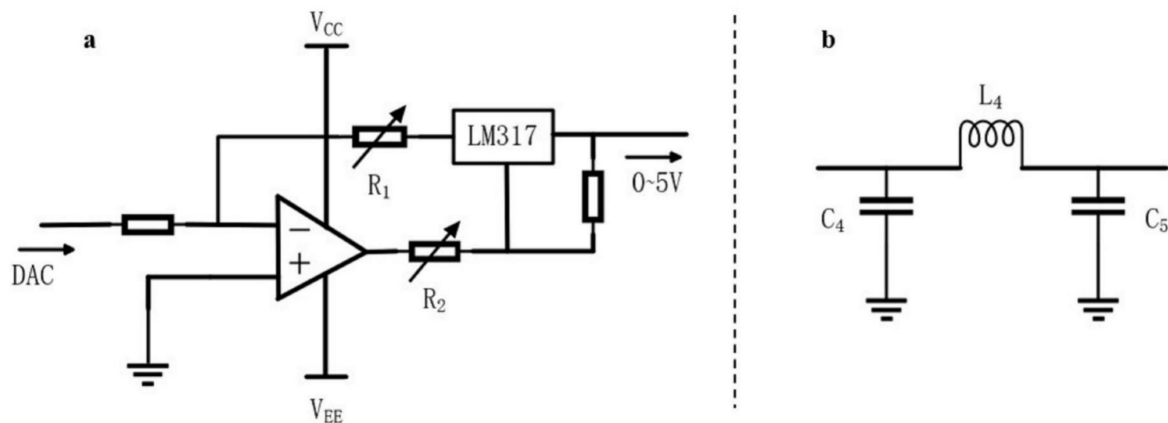
$$f_e = \frac{1}{2\pi\sqrt{L_3C_3}} \tag{10}$$

### 2.3. Dynamic Analysis

The dynamic analysis of this system mainly includes two aspects: (1) to realize the dynamic change of the supply voltage to the Class E PA from 0 to 5 V; (2) the modulation of  $V_g$  (duty ratio 1–50% and stimulation frequency 1–20 Hz). Based on the above requirements, the SCM STM32F103ZET6 development board is used as the core control board of the system.

#### 2.3.1. DAC Output

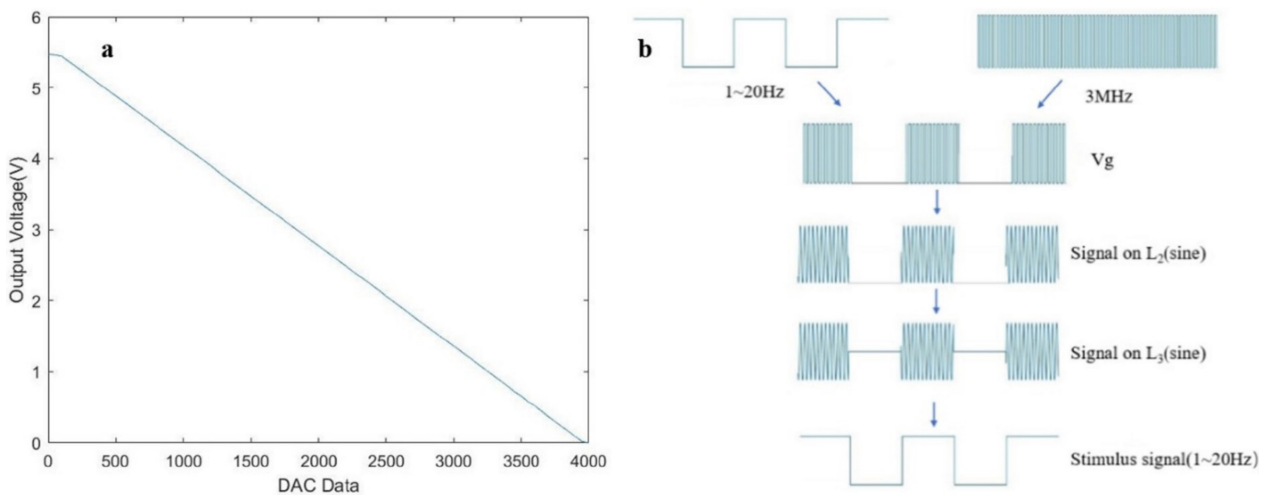
For convenience, we use the DAC output function of STM32 to obtain the controllable dynamic voltage with the target of 0–5 V. The specific circuit is shown in Figure 4a. And the Figure 4b shows the topology of the impedance matching network we used.



**Figure 4.** (a) The control circuit diagram of DAC output. The input is the DAC voltage of 0–3.3 V. The output voltage is 0–5 V; (b) The  $\pi$  type impedance matching network.

The output voltage of LM317 is controlled by adjusting  $R_1$  and  $R_2$ . Finally, the 0–5 V output voltage is controlled by 0–3.3 V DAC voltage in STM32. The  $V_{CC}$  and  $V_{EE}$  are powered by a jMD35-D18 AC-DC switching power supply. The DAC of STM32 is 12 bits, and the value can be set in the range of 0–4096 to obtain the DAC output voltage of 0–3.3 V (Figure 5a). The following formula fits the curve in Figure 5a.

$$Y = 3975 - 0.714U \tag{11}$$



**Figure 5.** (a) The relationship between the DAC setting values and the output voltages; (b) The signal transmission process.

$Y$  is the DAC data, and  $U$  is the actual output voltage (V). Here we achieve a 0–5 V power supply of the system, and the adjustment accuracy is about 0.7 mV.

### 2.3.2. The Signal Modulation

Generally, the electrical stimulation signals of nerve repair in rats are unfixed. We use the PWM (Pulse Width Modulation) output function of STM32 to conduct the signal modulation. In this study, TIM2 and TIM3 timers are selected to generate a high-frequency carrier and low-frequency stimulation pulse, respectively. The frequency calculation formula is:

$$f = \frac{72\text{MHz}}{(\text{TIM\_Period} + 1)/(\text{TIM\_Prescaler} + 1)} \quad (12)$$

The TIM\_Period and TIM\_Prescaler are adjustable parameters of the timers and are non-negative values. The pulse with different frequencies can be obtained by setting different parameter values (Table 3). Only a few frequencies are shown in the table. The frequency range can be expanded by setting parameters as required. The duty cycle can be controlled by comparison registers (TIMx\_CCR2–3) in STM32.

**Table 3.** The parameter design of the PWM output.

	TIM_Period	TIM_Prescaler	$f$ (Hz)	Note
1	23	0	$3 \times 10^6$	carrier pulse
2	7200	500	20	stimulation pulse
3	7200	1000	10	stimulation pulse
4	7200	2000	5	stimulation pulse

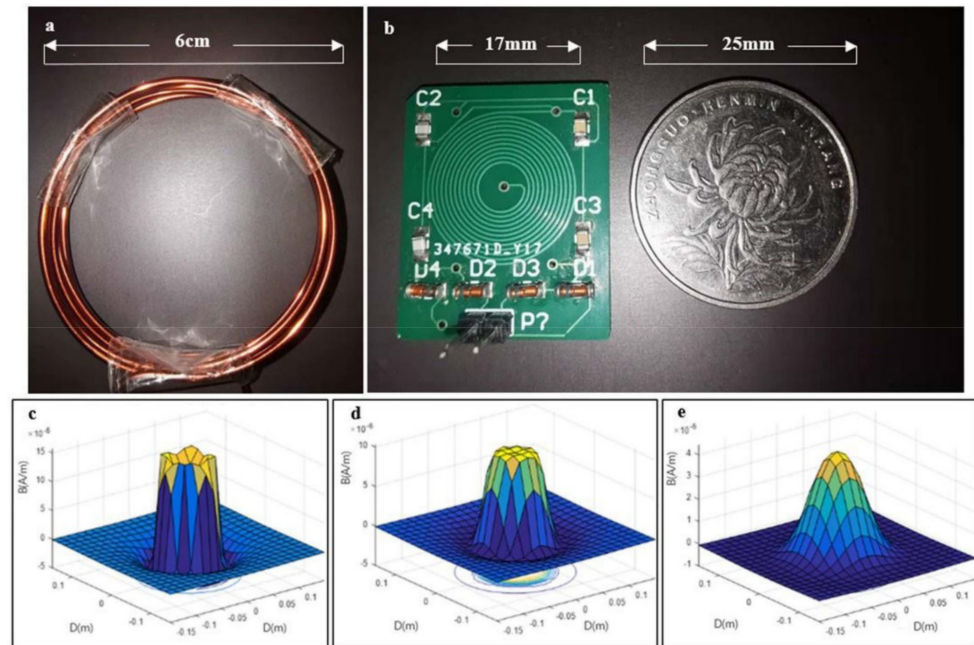
Figure 5b shows the signal transmission process. The high-frequency carrier pulse and the low-frequency stimulation pulse are modulated with the AND gate to obtain the final  $V_g$  signal. The  $V_g$  signal oscillates through Class E PA to obtain the signal on  $L_2$ . After that, the signal on  $L_3$  was obtained through wireless transmission. Finally, the low-frequency stimulation pulse of 1–20 Hz for neural electrical stimulation was generated through demodulation. Moreover, according to actual needs, two-timer parameters can be set to obtain a larger frequency of the stimulation pulse.

So far, the characteristic parameters of the wireless stimulator designed in this study have adjustable ranges, that is, a frequency of (1–20 Hz) and a duty cycle of (1–50%).

### 3. Results

#### 3.1. The Fabrication of Coil

Figure 6a,b show images of the  $T_X$  coil and the  $R_X$  coil, respectively.



**Figure 6.** The fabrication of coils: (a) The  $T_X$  coil; (b) The  $R_X$  coil and receiving circuit on the left; (c) The magnetic field stimulation of  $T_X$  coil on the plane with transmission distance  $D = 1$  cm; (d)  $D = 2$  cm; (e)  $D = 5$  cm.

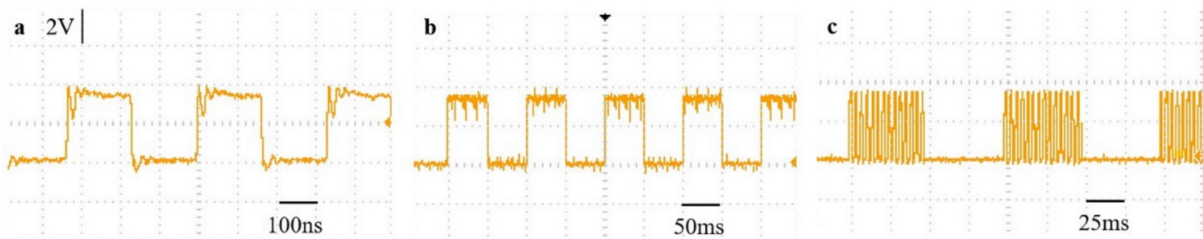
The  $T_X$  coil made of 1.5 mm-enameled wire is a cylindrical spiral coil, with a diameter of about 6 cm and 6 turns. The resistance of the  $T_X$  coil is about  $0.5 \Omega$ , and the inductance is  $3.68 \mu\text{H}$ . Because the  $T_X$  coil is in the external circuit, the design has high flexibility and variability and can be manufactured in the laboratory according to actual needs. In this study, the  $T_X$  coil is used for neural electrical stimulation in rats, and the coupling distance is generally 1–4 cm. Figure 6c–e shows the magnetic field stimulation results of the  $T_X$  coil under MATLAB2018a.

The  $R_X$  coil is a planar helical coil that is mounted on a four-layer FR-4 PCB of  $20 \text{ mm} \times 25 \text{ mm}$  in size with the receiving circuit. The outer diameter is about 1.7 cm. The diameter of wire and the spacing between the wires are about 0.25 mm. And the turn-number of the coil is twelve. The resonant capacitance of the receiving circuit is about 494 pF, and  $L_3 = 5.37 \mu\text{H}$  is calculated from (9). The inductance is measured to be  $5.59 \mu\text{H}$  using the UT612 LCR digital bridge. The actual resonant frequency  $f_e = 3.03 \text{ MHz}$  is calculated by the Formula (10).

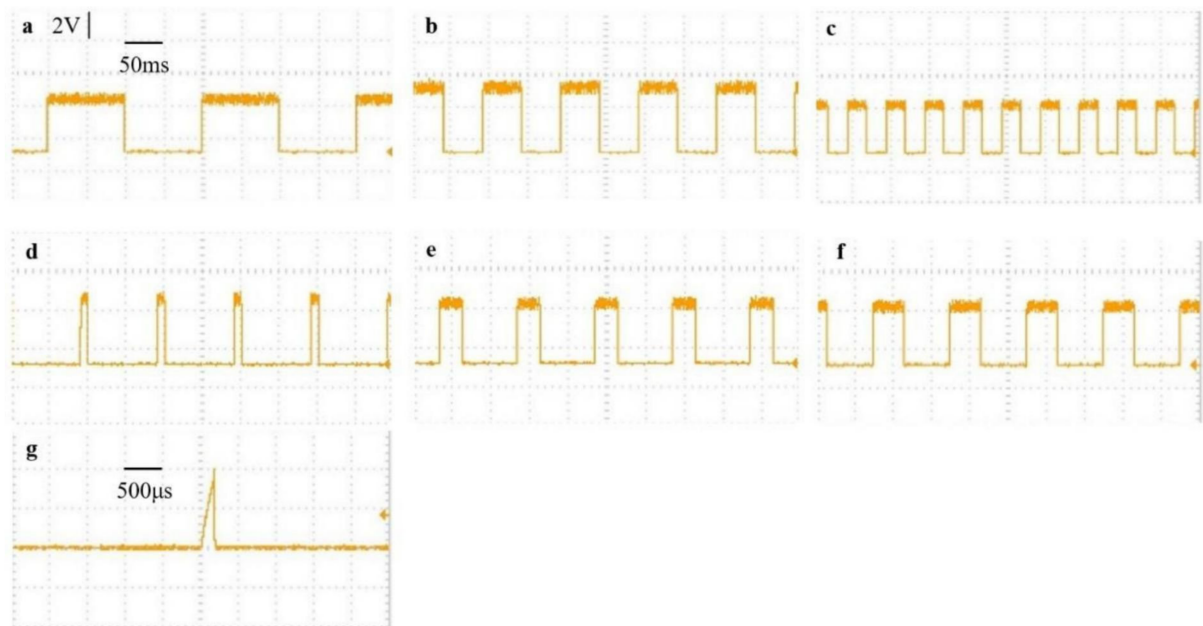
#### 3.2. Signal Transmission Experiment

In the signal transmission process of this study, 1–20 Hz stimulation pulse ( $S_1$ ) and 3 MHz carrier pulse ( $S_2$ ) are generated by STM32. The gate drive signal  $V_g$  is modulated by these two signals using the AND gate. In 10 Hz, for example, Figure 7 shows the  $V_g$  signal generation process. After that,  $V_g$  generates high-frequency sinusoidal signals at  $L_2$  and  $L_3$  after oscillating through class E PA. The signal received on  $L_3$  passes through the rectifier to obtain the final stimulus signal  $S$ . As shown in Figure 8a–f, we get stimulus signals with different frequencies and duty cycles. Moreover, If the duty cycle of  $S$  is 1%, we can get the minimum pulse width of the stimulus signal. Figure 8g shows that the minimum pulse width is about 200  $\mu\text{s}$ .





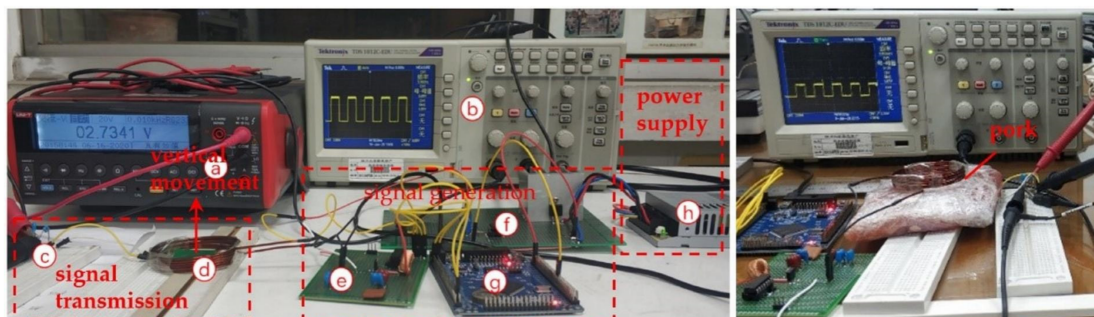
**Figure 7.** The modulation of the  $V_g$  signal (a) the high-frequency carrier pulse  $S_2 = 3$  MHz; (b) the low-frequency stimulation pulse  $S_1 = 10$  Hz; (c) the  $V_g$  signal.



**Figure 8.** The stimulus signal of different parameters: (a–c) the stimulus signal of different frequencies: 20 Hz, 10 Hz, 5 Hz; (d–f) the stimulus signal of different duty cycles:10%, 20%, 40%; (g) the minimum pulse width is about 200  $\mu$ s when the duty cycle is 0.01.

### 3.3. Output Performance Test

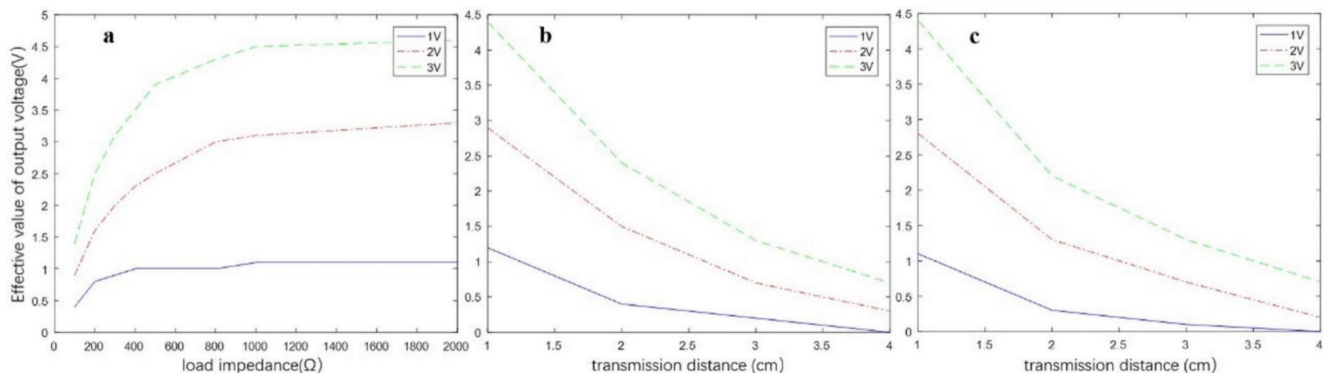
Figure 9 on the left shows the instrument and circuit used for the test experiment. Figure 9 on the right shows the power transmission experiment by using pork as the transmission medium.



**Figure 9.** The experiments of output performance test with different media: air (left) and pork (right). The testing instruments used are multimeter (UNI-T UT805A, a), oscilloscope (Tektronix TDS 1012C-EDU, b). In the picture, the c is the simulated load impedance; the d represents for the  $T_x$  coil and the  $R_x$  coil and they move vertically in a coaxial way; the e is the Class E PA circuit; the f is LM317 controllable constant-voltage circuit; the g is STM32 microcontroller; and the h is switching power supply to make the LM317 work.

We carried out an output capacity test, distance test, and medium test on the designed system, on the conditions that  $S_1 = 3$  MHz,  $S_2 = 10$  Hz, and duty cycle = 0.5.

First, at transmission distance  $D = 1$  cm, the output capacity of the system is tested by changing the value of load impedance ( $100 \Omega$ – $2$  k $\Omega$ ) and VDD value (1 V, 2 V, 3 V). The voltage effective values under different loads and supply voltages are shown in Figure 10a. After that, the load impedance value (1 k $\Omega$ ) was fixed to test the variation of the induced voltage with the transmission distance (Figure 10b). In the end, we repeated the experiment using 2 cm thick pork as the medium (Figure 10c).



**Figure 10.** The results of output performance test: (a) the voltage effective values under different loads and supply voltages; (b) the transmission distance test in the air; (c) the transmission distance test with the 2 cm thick pork.

#### 4. Discussion

In this section, we describe the development and testing of a fully implantable wireless nerve stimulator for neuroelectrical stimulation rehabilitation experiments. The device is divided into an external circuit and an implantable part through a modular design that facilitates its miniaturization. In the process of device development, we used technologies including, but not limited to, single-chip microcomputer technology, signal modulation, and demodulation, wireless power supply, and integration/packaging. The system can realize flexible control of stimulus parameters to adapt to different application scenarios.

In this study, electrical stimulators were designed to target small animals such as rats. We designed the transmission distance between 1 and 4 cm, using an electromagnetic field frequency of 3 MHz. Through the approach of energy transmission, the system can meet the conditions required by neural electrical stimulation within the transmission distance. For the wireless energy transmission of electromagnetic waves, the overall trend is that the transmission efficiency increases with the increase of working frequency [32]. However, the increased working frequency can also cause the biological tissue to absorb too much energy, which can lead to tissue overheating and cause safety problems such as burns. Accordingly, we set 3 MHz as the operating frequency of the system.

Under these conditions, we discuss the experimental results as follows.

As shown in Figure 6, both transmitting and receiving coils can be manufactured in the laboratory with convenient features. The transmitting coil can be made at any time according to the desired inductance parameters. The outer diameter of the receiving coil is only 1.7 cm, which can meet the requirements of miniaturization and is implantable. Without changing parameters such as inductance, the diameter of the  $R_X$  can be reduced by increasing the layer of coils, thereby reducing the size of the stimulator. However, more layers increase the resonant impedance of the coil. The diameter  $D_0$  is limited by the wire diameter  $w$ , the spacing between each turn  $b$ , and the number of turns  $N$ . Therefore, in our work, the  $R_X$  coil is, at most, 4 layers with a minimum diameter of about 8 mm. The magnetic field simulation analysis of the transmitting coil shows that the magnetic field intensity around the transmitting coil is on the order of  $10^{-6}$ , which belongs to the safe range [33].

It can be seen from Figure 7 that the  $V_g$  signal will be directly used for switching control of the transistor, thus generating the oscillation wave patterns transmitting coil and sending them to the receiving circuit. Figure 8 shows that the minimum pulse width of  $S$  is about 200  $\mu\text{s}$ . Duty cycle (1–50%), and frequency (1–20 Hz) can be changed within the set range, with an accuracy of 1% and 1 Hz, respectively. Simultaneously, the DAC function of the SCM controls the input voltage of the Class E PA in 0–5 V, and the adjustment accuracy is 0.7 mV. The stimulator can perform diverse and flexible work according to actual needs.

As can be seen from Figure 10a, when the resistance value of load impedance is large ( $>800 \Omega$ ), the induced voltage at both ends of the load is stable, and when the power supply is 3 V and the load impedance is 800  $\Omega$ , the effective value of the induced voltage is 4.3 V. It can be concluded that the peak voltage of square wave at this time is about 8.6 V, which is greater than the required stimulation voltage (5 V). Therefore, the wireless stimulator can effectively stimulate the nerves of rats. When the resistance value of the load impedance is small, the main reason for voltage reduction is that the wireless transmission power is not enough to meet the power consumption, resulting in a certain voltage drop.

Figure 10b shows that the induced voltage of the load varies significantly with the distance between the coils. Moreover, the further the distance, the lower the induced voltage value, which is because the coupling coefficient between the coils decreases as the distance is increased. In practical applications, if the receiving coil is embedded in rats and the transmitting coil is used for power transmission in vitro, the transmission distance is generally less than 2 cm. When the supply voltage is 3 V, the effective value of the induced voltage can reach 2.3 V at the transmission distance of 2 cm, that is to say, the voltage of the induced waveform can reach about 4.6 V. Additionally, the effective value is 0.7 V at the transmission distance of 4 cm, and effective electrical stimulation can be made.

In addition, the comparison between Figure 10b,c shows that the induced voltage of the load at the pork medium is very similar to that of air. The main reason for the difference is the interval error of the transceiver coil and the deviation of the axis of the transceiver coil.

However, some limitations should be noted. First, the induced voltage varies obviously with distance, and the coupling distance is small. Second, the study did not consider the biocompatibility of the implants. Third, the power of the stimulus needs to be improved. Fourth, this study only designed stimulation tools without animal experiments. Finally, biodegradable preparation methods of inductors, capacitors, and other components have been realized [34,35], and biodegradable components can be used to avoid secondary trauma in future implanted circuit design.

In summary, we developed a wireless stimulator for rats. Given the size of the stimulator, it will be used primarily to stimulate the sciatic nerve in rats, with the spinal cord and brain also being potential targets. Both stainless steel electrodes [36] and Mg electrodes [5] can be used in experiments. It has been concluded that subcutaneous implantation of a 25-g ( $17.5 \text{ cm}^3$ ) telemeter in a 170 g male rat does not harm animal health and well-being [37]. The stimulator we designed is much smaller, so the normal behavior of the rats will not be affected. Moreover, to minimize physical damage to tissue, a flexible receiver coil and flexible polyimide printed circuit board will be used in animal experiments. Koo, J. et al. [5] demonstrated that the flexible polyimide printed circuit can better fit neural tissue, and it can also provide a stable, low-profile, non-invasive means of facilitating chronic neural stimulation.

## 5. Conclusions

In this paper, a wireless stimulator is designed and tested for electrical stimulation. The stimulator consists of a transmitting circuit and an implanted circuit. At a distance of 1–4 cm and a frequency of 3 MHz, the magnetically coupled resonant wireless power transfer (MCR-WPT) technology is used to transmit energy. The transmitter uses STM32F103ZET6 SCM as the signal control part and the Class E PA as the power amplification part. The receiver is a circuit built with a passive device that can be implanted in rats. After signal

transmission experiments and output performance tests, the stimulator can output a single-phase pulse signal with variable parameters, and the power is about 25 mW. The stimulation frequency varied from 1 Hz to 20 Hz, the duty cycle varied from 1 to 50%, and the minimum pulse width was 200  $\mu$ s. The input voltage of the stimulator is variable from 0 to 5 V. When the input voltage is 3 V and the duty cycle is 0.5, the effective value reaches 4.3 V, 2.3 V, and 0.7 V at the transmission distance of 1, 2, and 4 cm, respectively. Therefore, the stimulator can perform low frequency, safe and controllable wireless stimulation over a short distance.

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